INSTRUCTION BOOK

for

MODEL 320A
TRUE ROOT MEAN SQUARE
ELECTRONIC VOLTMETER

BALLANTINE LABORATORIES, INC.
BOONTON, NEW JERSEY

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INSTRUCTION BOOK CONTENTS

FOR

MODEL 320A TRUE ROOT MEAN SQUARE

ELECTRONIC VOLTMETER

		Page
1.	**GENERAL INFORMATION**	1
1.1	Purpose and Use	1
1.2	Some Applications of the Model 320A	1
1.3	Technical Characteristics	1
2.	**OPERATION**	2
2.1	Power Connection	2
2.1.1	Line Voltage Conversion	2
2.2	Starting Procedure	2
2.2.1	Warmup Period	3
2.3	Function of Controls	3
2.3.1	Connectors and Terminals	3
2.3.2	Controls	3
2.4	Voltage Measurement	3
2.4.1	Scale Adjustment	3
2.5	Possible Error Sources	4
2.5.1	Pointer Stays Extreme Left	4
2.5.2	Indication is High	4
2.5.3	Pointer Stays Extreme Right	4
2.5.4	Input Impedance and Loading Error	4
2.5.5	Ground Current Error	4
2.5.6	Transmission Line Error	4
2.5.7	Magnetic and Electrostatic Fields	4
2.6	Use of the DB Scale	4
2.7	Effect of Harmonics	5
2.8	Effect of High Crest Factor	5
2.8.1	Flat Top Waveforms	5
2.9	Effect of the DC Component	6
2.10	Measurement of Noise and Random Signals	6
2.11	Measurement of Waveforms with Higher Crest Factor than 5	6
2.12	Measurement on .3 mV Range	7
2.13	Measurement of Voltages Above 300 V	7
2.14	Use as Null Detector	7
2.14.1	To Use the Model 320A as Null Detector	7
2.14.2	For Approximate Voltage Measurements Below 100 μV	7
2.15	Overload Characteristics	7
2.16	Influence of Power Line Voltage	7
2.17	RMS Current Measurements	8
2.18	Power Dissipation Measurement	8
2.19	Use of Mean Square Output	8
2.20	Use as Amplifier	8
3.	**CIRCUIT DESCRIPTION**	9
3.1	Input and Midsection Attenuators	9
3.1.1	Input Attenuator	9
3.1.2	Midsection Attenuator	9
3.2	Input Amplifier	10
3.3	Output Amplifier	10

Page

3.4 Phase Inverter 10
3.5 Square Law Detector 10
3.6 Power Supply 10

4. MAINTENANCE 10
4.1 General 10
4.2 Recommended Maintenance 11
4.2.1 Meter screw adjustment 11
4.2.2 1,000 Hour Check 11
4.2.3 2,000 Hour Check 11
4.3 Necessary Maintenance Equipment 11
4.3.1 DC Voltmeter 11
4.3.2 Stable AC Voltage Source 11
4.3.3 Calibrated AC Voltage Source 11
4.3.4 Tube Tester 11
4.3.5 Calibrated, Variable RF Voltage Source 11
4.3.6 Variable Metered AC Power Supply 11
4.4 Removal of the Case 11
4.5 Scale Adjustment 11
4.6 1,000 Hour Check 11
4.6.1 Tests Comprising the 1,000 Hour Check 11
4.6.2 Bias Test of V7 Tube 11
4.6.3 Meter Screw Adjustment 11
4.6.4 Line Voltage Stability Test 11
4.6.5 Midband Accuracy Test 11
4.7 2,000 Hour Check 12
4.7.1 Tests Comprising the 2,000 Hour Check 12
4.7.2 Power Supply Test 12
4.7.3 Bias Test on V7 12
4.7.4 Tube Test 12
4.7.5 Frequency Response Test 12
4.7.6 Midband Accuracy Test 13

5. SERVICE AND TROUBLESHOOTING 13
5.1 General 13
5.2 Necessary Equipment 13
5.2.1 DC Voltmeter, Ohmmeter 13
5.2.2 Sensitive DC Voltmeter 13
5.2.3 Sensitive AC Voltmeter 13
5.2.4 Signal Generator 13
5.2.5 Accurate RF Voltage Calibrator 13
5.2.6 Accurate Calibrated 1 kc Voltage Source 13
5.2.7 HF Diode Voltmeter 13
5.3 Simple Service Problems 13
5.3.1 Removal of the Case 13
5.3.2 Pilot Light Replacement 13
5.3.3 Fuse Replacement 13
5.3.4 Tube Replacement 13
5.4 DC Voltage Check 13
5.5 Signal Tracing 14
5.6 Check of Square Law Detector and Indicator 14
5.6.1 Continuity Test 14
5.6.2 Diode Test 14
5.6.3 Squarer Bias Test 14
5.6.4 Electrical Linearity 15
5.7 Frequency Response Alignment 15
5.7.1 Low Frequency Response 15

Page

5.7.2 High Frequency Response 16
5.7.3 Amplifier Response Alignment 16
5.7.4 0.3 mV Range Adjustment 16
5.7.5 C30 Midsection Attenuator Adjustment 16
5.7.6 C31 Midsection Attenuator Adjustment 17
5.7.7 Input Attenuator Adjustments 17
5.8 Sensitivity Adjustment 18
5.8.1 R71 Sensitivity Adjustment 18
5.8.2 R29 .3 mV Range Sensitivity Adjustment 18
5.9 Table of Calibration Controls 18
5.10 Service of Attenuators 18
5.11 Intrinsic Noise and Hum Test 18
5.12 Service of the Power Supply 18
5.13 Troubleshooting Chart 19

6. SHIPPING INSTRUCTIONS 20

7. REPLACEMENT PARTS LIST 25

ILLUSTRATIONS AND DIAGRAMS

Figure 1. Ballantine Model 320A True RMS Voltmeter IV
Figure 2. Rack Panel Mounted Ballantine Model 320A-S/2 True RMS Voltmeter IV
Figure 3. Input Resistance Change with Frequency 4
Figure 4. Duty Cycle and Crest Factor of a Pulse Train 5
Figure 5. Connection of the External Blocking Capacitor 6
Figure 6. Crest Factor Error with White Noise 6
Figure 7. Use of External Attenuator to Extend the Crest Factor Limit 6
Figure 8. Calibration of 0.3 mV Range on Null Detector Mode 7
Figure 9. Simplified Circuit Diagram 9
Figure 10. Connection of Micropotentiometer 12
Figure 11. Typical Frequency Response 15
Figure 12. Connection of HF Transfer Voltmeter 17
Figure 13. Side View of Model 320A (right) Showing Location of Tubes and Adjustments 22
Figure 14. Side View of Model 320A (left) Showing Location of Tubes and Adjustments 23
Figure 15. Front View of Model 320A (escutcheon plate removed) Showing Location of Adjustments 24
Figure 16. Schematic of Model 320A True RMS Voltmeter ..... Rear

TABLES

Table I DB Level Conversion 5
Table II Maximum Correction for Harmonics 5
Table III Corrections on 0.3 mV Range 7
Table IV Effect of Line Voltage 8
Table V Attenuation Ratios 9
Table VI Grid Voltage Limits 14
Table VII Squarer Bias Limits 15
Table VIII Response Alignment Controls 15
Table IX Calibration Table 21

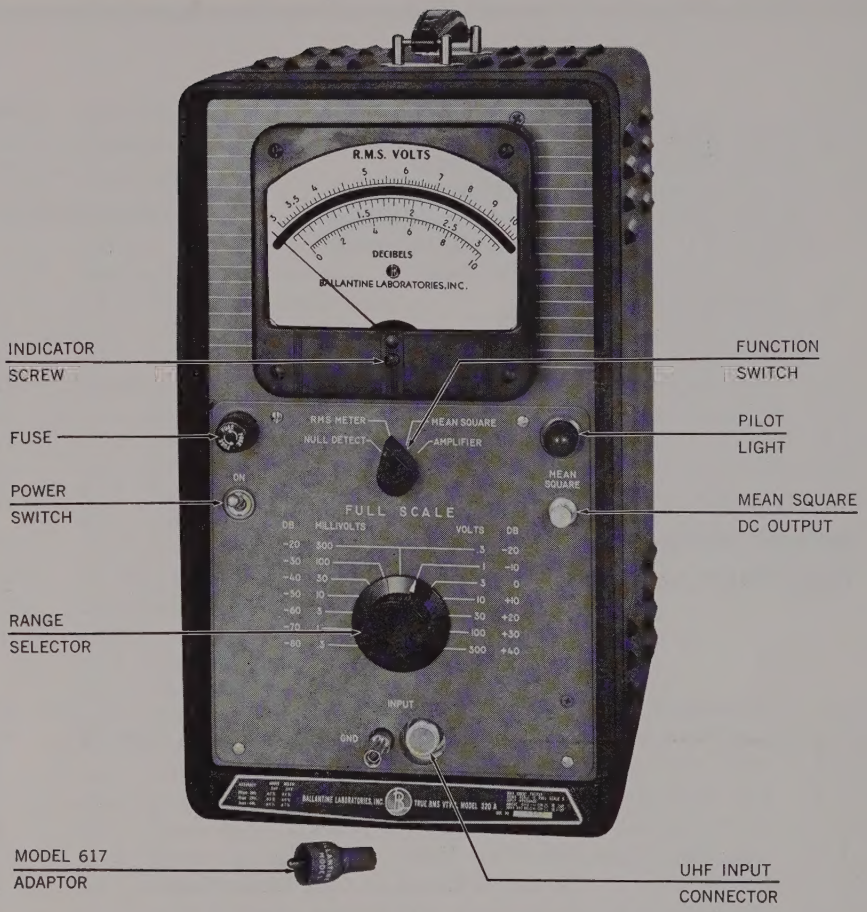


Fig. 1. Ballantine Model 320A True RMS Voltmeter

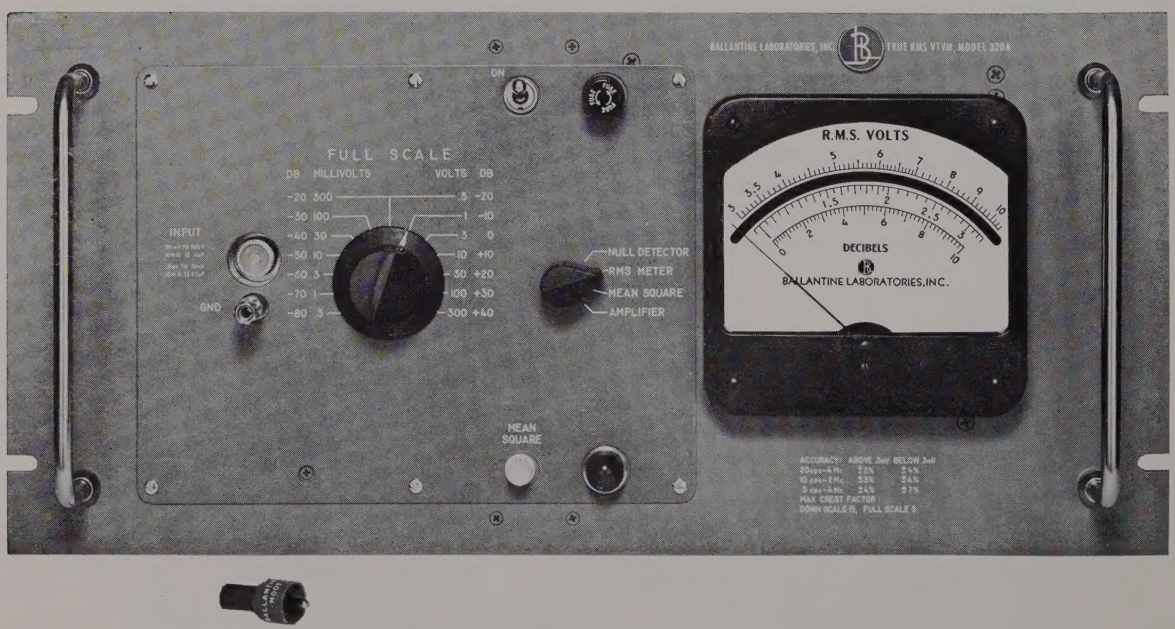


Fig. 2. Rack Panel Mounted Ballantine Model 320A-S/2 True RMS Voltmeter

1. GENERAL INFORMATION

1.1 Purpose and Use

The Ballantine Model 320A is a sensitive, wide band, TRUE ROOT MEAN SQUARE responding ac voltmeter with frequency range of 5 cps to 4 Mc, voltage range of 100 μV to 330 V and a basic accuracy of better than 2%.

Its main characteristics — accuracy, band width and maximum crest factor — are chosen so that the instrument is suitable for accurate measurement of a wide variety of ac waveforms which are encountered in current acoustical, vibrational, noise, pulse and magnetic measurements. Besides sinewaves it measures distorted waveforms, random noise and pulse trains with duty cycles as low as 0.004 and pulse widths as short as 0.3 μsec.

The true rms response of the Model 320A is achieved by means of the Ballantine patented square law detector which is an accurate, wide band, segmented function generator. In conjunction with a full wave rectifier, this detector provides a precise, wide dynamic range, and instantaneous squaring action which is unaffected by the normal range of ambient temperature, overload and component aging. The indication is achieved on two 4.5" long, mirrorbacked, logarithmic scales. *The same precision and accuracy applies to measurements at any point on the scales.* In fact, higher accuracy specification at full scale indication is possible, but it is Ballantine's philosophy that specifying higher accuracy for only certain parts of the scale does not increase the usefulness of the voltmeter. An associated linear db scale with dbv (1 V) as reference is supplied with the standard instrument.

In the design of the voltmeter special consideration is given to operation under disturbing influences. The coaxial type UHF input terminal provides a low ground impedance input which reduces the error due to ground currents and allows the use of coaxial input for accurate high frequency measurements. A Model 617 adapter supplied with the voltmeter converts the coaxial input to standard ¾" spacing binding posts, an input preferred for low frequency measurements. The warmup drift is kept small, so that the voltmeter is usable 20 sec after being turned on. The effect of line voltage transients is minimized so that the Model 320A may be used under very unfavorable line conditions.

Besides the excellent initial performance, great emphasis is placed on the reliability and life of the voltmeter. Because of the conservative loading of tubes and components, proper amount of feedback and the reserve in accuracy specifications, it is expected that the great majority of voltmeters will meet its specifications over 5,000 hours of normal laboratory use without the need for readjustments or recalibration. The voltmeter design is such that whenever a tube change may be necessary it does not influence the voltmeter performance over most of the useful band and does not require extensive recalibration or adjustment. The Model 320A stability against line voltage changes is such that the line voltage variation from 105 to 125 V will not set the instrument out of specifications. These features make the Model 320A a very economical instrument with respect to service and upkeep which, considering its low initial cost, adds more to the value of the voltmeter.

Beyond its basic function as a voltmeter, the Model 320A may be used as a null detector with a sensitivity down to 10 μV. It also can be used as a wide band amplifier with floating, grounded or symmetrical output. Up to 50 V rms may be obtained from the amplifier output. A dc current proportional to the mean square value of input voltage may be obtained, to drive a recorder or other readout device. By means of Ballantine series 600 calibrated shunts, the Model 320A may be converted to a true rms responding ac ammeter with sensitivity down to 0.1 μA.

The construction of the voltmeter uses high grade printed wiring boards which guarantee uniformity and easy serviceability of the instrument. Modern frame grid tubes with highest grade components are used throughout the circuit. The design is based on 7 years experience with the Ballantine Model 320 — an accepted standard of rms voltmeters throughout the industry.

The Model 320A is available in portable or rack mounted version for 115 V or 230 V, 50 to 420 cps power line.

1.2 Some Applications of the Model 320A

General rms voltage measurement in grounded circuits

Noise figure determination of tubes, transistors, networks and resistors

True rms vibration measurements, in conjunction with vibration pickup

White noise tests of audio equipment and noise level checks in wide band carrier and video systems

Mean square sound pressure measurement and acoustic studies with suitable microphones

As audio output power meter with appropriate load impedance

Measurement of total harmonic distortion in conjunction with a fundamental suppression filter

1.3 Technical Characteristics

Voltage Range	0.1 mV to 330 V in 13 ranges 10 μV sensitivity as null detector
Frequency Range	5 cps to 4 Mc
Waveforms	Sine, complex, pulse or random
Max Crest Factor	15 for down scale indication 5 for full scale indication

Accuracy at ANY Point on the Scale

Sine, complex, and random waveforms

For voltages above 0.3 mV

20 cps to 400 kc	±2% of indication
10 cps to 2 Mc	±3% of indication
5 cps to 4 Mc	±4% of indication

For voltages from 0.1 mV to 0.3 mV

10 cps to 2 Mc	±4%
5 cps to 4 Mc	±7%

Pulse trains and flat top waveforms

Crest factor below 4 (FS) to 10 (DS) as above

Crest factor below 5 (FS) to 15 (DS) ±2% additional uncertainty

Scales

Voltage scales:	0.95 to 3.3, logarithmic 3.0 to 10.6, logarithmic separated with mirror
db scale:	0 - 10 db linear

DB Range —80 to +40 db referred to 1 V

Input Impedance

0.3 mV to 10 mV range	10 MΩ parallel with 27 pF
30 mV to 300 V range	10 MΩ parallel with 11 pF

Equivalent Input Noise

Input shorted, 0.3 mV range	20 μV max
Input open, 0.3 mV range	40 μV max

Response Time

Indicator	1 to 2 sec
Mean square output	0.1 sec

Warmup Time

Usable after	20 sec
Less than 0.3% drift after	15 min

Ambient Temperature Effect

Instrument calibrated at 25°C

Drift less than 1/10 of stated accuracy over 20°C to 30°C

Stability

Expected calibration stability	5000 operating hrs
Recommended calibration periods	1000 operating hrs
Effect of tube replacement	minor effect above 1 Mc only

Amplifier Characteristics

Output	Binding posts in rear, balanced or single ended
Max gain	90 db ±1 db adjustable in 10 db steps with range selector
Max output	50 V into 20,000 ohms 10 V into 2,000 ohms
Source impedance	350 Ω at 1 kc
Frequency response	±0.5 db from 5 cps to 2 Mc —1 db at 4 Mc, 25 pF max loading

Mean Square Output

Output voltage	—0.2 V for full scale deflection
Max output	—1.0 V
Source resistance	1,000 Ω

Power Requirements

Voltage	105 to 125 V (210 to 250 V)
Frequency	50 cps to 420 cps
Power consumption	90 W
Fuse	115 V — 0.8 A Slo-blo 230 V — 0.5 A Slo-blo
Effect of line voltage change	±10 V (±20 V) line voltage change affects the sensitivity by less than 1/10 of guaranteed accuracy

Vacuum Tubes and Semiconductors

Tubes	2 — 5654 1 — 12B4A 1 — 5V4G 4 — 6EJ7 1 — OA2 1 — 5651 1 — 8223 or 1 — 6DJ8 1 — OB2
Semiconductors	10 — germanium diodes 1 — silicon diodes 2 — silicon rectifiers

Mechanical Specifications

Color	Portable: Gray panel, black wrinkle case Rack: Gray or special to customer's specification
Dimensions	Portable: 15⅛" H x 8⅜" W x 11¾" D Rack: 8-23/32" H x 19" W x 12¾" D
Weight	Portable or rack: 22 lbs. Shipping wgt.: Portable 27 lbs., rack 40 lbs.

2. OPERATION

2.1 Power Connection

The voltmeter is supplied ready to operate on 105 - 125 V, 50 cps - 420 cps or 210 - 250 V, 50 cps - 420 cps power line, as indicated on the decal mounted adjacent to the power cord.

2.1.1 Line Voltage Conversion from 115 V to 230 V or vice versa is possible. For conversion from 115 V operation to 230 V operation, connect the power transformer leads as follows:

Disconnect black-red lead from terminal 4 and black-yellow lead from terminal 6, and connect both together to terminal 5. Replace fuse F1 0.8 ampere by 0.5 ampere Slo-blo fuse. Detailed information for 115 V/230 V conversion is given on the schematic at the end of this instruction book.

2.2 Starting Procedure

Insert the three prong plug into a proper ac power outlet. Use conversion unit where only two prong outlets are available. Move the power switch to ON position. The pilot light should glow. The instrument is ready for use after a short warmup period.

2.2.1 Warmup Period

The instrument is usable after approximately 20 seconds. After 15 minutes the indication will be within 0.3% of the final value.

When the instrument has not been in use for many months or has been stored in high humidity allow a warmup period of at least one hour.

NO SOURCE OF HEAT SHOULD BE CLOSE TO THE BACK OF THE INSTRUMENT. PERMIT AIR TO CIRCULATE FREELY AROUND THE VOLTMETER TO AVOID EXCESSIVE TEMPERATURE RISE IN THE INSTRUMENT.

The voltmeter is calibrated in the vertical position. For convenience in reading, a tilting device is provided beneath the case.

2.3 Function of Controls (See Figure No. 1 for control locations)

Power Switch ON — Turns the instrument on.

Function Switch

NULL DETECT — The instrument operates as indicator for voltages normally below the voltmeter scale.

RMS METER — Instrument operates as voltmeter.

MEAN SQUARE — Instrument operates as an ac to dc converter. The dc output proportional to the square of ac input voltage is available from MEAN SQUARE connector.

AMPLIFIER — The instrument operates as wide band amplifier with the output terminals at the rear of the case.

Range Selector

FULL SCALE .3 to 300 mV .3 to 300 V — Selects the voltage range in RMS METER or NULL DETECT operating mode. Changes the amplifier gain in 10 db steps in AMPLIFIER mode or dc mean square output in 20 db steps in MEAN SQUARE mode of operation.

2.3.1 Connectors and Terminals

INPUT — Type UHF connector for signal input. May be converted to binding posts with Ballantine Model 617 adapter.

GND — Ground binding post connected to the Voltmeter chassis and ground prong of the line cord.

MEAN SQUARE — Type BNC connector for dc output in MEAN SQUARE mode of operation.

AMP OUTPUT — Four binding posts for floating, symmetrical or single ended amplifier output in AMPLIFIER mode of operation.

2.3.2 Controls — adjustable by screwdriver through the holes in the rear of the voltmeter.

LIN ADJ — Adjust electrically the down scale sensitivity in RMS METER and MEAN SQUARE mode of operation.

NULL ADJ — Adjust the rest position of the pointer in NULL DETECT mode of operation.

SENS ADJ — Adjust the instrument sensitivity in RMS METER, NULL DETECT and MEAN SQUARE modes of operation.

At the front of the instrument —

Indicator Screw — Adjust the down scale indication in RMS METER mode of operation.

2.4 Voltage Measurement

For true rms ac voltage measurement:

Set	To
Function Switch	RMS METER
Range Selector	300 V

Connect the unknown voltage to the input and rotate the range selector counterclockwise until indication appears on the meter scale. Take reading from the scale indicated by the range selector. If the indication is down scale, below the red db scale, select the next lower voltage range to bring the indication close to full scale for best readability. Because of the logarithmic meter scale and Ballantine special design, the *accuracy of the measurement is the same at any point on the scale,* with a logarithmic scale indicator the pointer rest position is normally off scale.

At approximately every 200 operating hours or when the measurement with utmost accuracy is desired, check the scale adjustment.

2.4.1 Scale Adjustment checks the scale linearity of the logarithmic indicator. It is equivalent to the zero adjustment in linear meter. For scale adjustment,

Set	To
Function Switch	RMS METER
Range Selector	1 V

Apply a stable 100 cps to 10 kc signal to the input. Adjust the amplitude to give an indication exactly on 10 on upper voltage scale. Rotate range selector to 3 V range. If the indication is not at 1.0 of lower voltage scale adjust the meter screw at lower center of the meter face until indication is 1.0 $\pm$0.2%. Repeat the procedure until full scale indication on 1 V range agrees exactly with down scale indication on 3 V range.

Do not use the meter screw to bring the pointer on marked scale when no signal is applied to the voltmeter

2.5 Possible Error Sources

The causes of some obvious measuring difficulties are listed below.

2.5.1 Pointer Stays Extreme Left when no signal is applied to the input, voltmeter is turned off or not connected to the power line, the input voltage is lower than the range selected, function switch is on MEAN SQUARE or AMPLIFIER position.

2.5.2 Indication Is High and the pointer stays on scale even after the next higher voltage range is selected. This happens when the function switch is in NULL DETECT position.

2.5.3 Pointer Stays Extreme Right when selected voltage range is too low or high hum is picked up by input circuitry.

Because of the wide frequency range and high sensitivity of the Model 320A, other measuring errors may be introduced when proper measuring techniques are not used. The most common error sources are listed below. A more comprehensive analysis of errors is given in NBS Conference on Standards and Electronic Measurement paper: Techniques and Errors in High Frequency Voltage Calibration by Dr. E. Uiga and W. F. White. A copy of the paper is available from Ballantine Laboratories without charge.

2.5.4 Input Impedance and Loading Error

The input impedance of the Model 320A depends on the setting of the Range Selector and the frequency of the signal measured. It may be represented with a parallel combination of input capacitance and resistance.

The input capacitance is 27 pF for ranges .3 mV to 10 mV and 11 pF for ranges 30 mV to 300 V. The variation of input capacitance with signal frequency is small.

The input resistance is 10 MΩ at low frequency. It changes with frequency as indicated in figure 3.

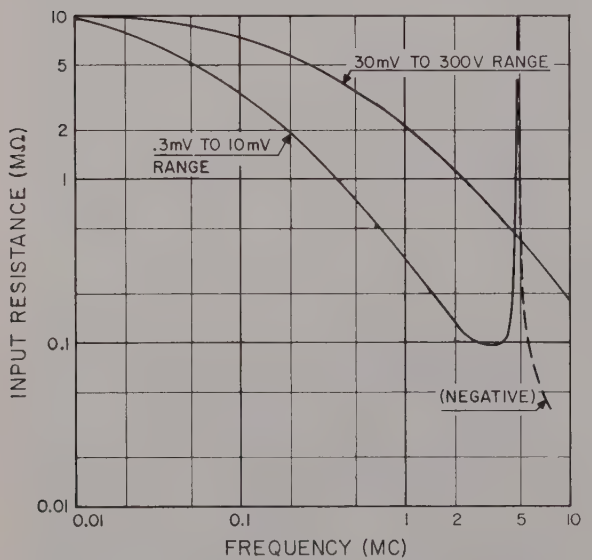


Fig. 3. Input Resistance Change with Frequency

As seen from the graph the input impedance on 0.3 mV to 10 mV ranges is negative at frequencies above 5 Mc.

This may cause an undamped resonance circuit tuned to 5 - 10 Mc to oscillate when voltmeter is connected across it. When that occurs, connect a parallel resistor with value equal to the negative resistance of input impedance across the input of the voltmeter. This resistor may be 100 kΩ or less, but should never be smaller than necessary to stop the oscillations.

The effect of the voltmeter input impedance and the impedance of the connecting leads is to load, damp or detune the circuit. This should be considered when high impedance sources above 100 kΩ or tuned circuits are measured.

2.5.5 Ground Current Error is caused by the ground currents of signal or power line frequency which may flow through the ground connection. This current develops a voltage drop in ground connection. This is measured together with the signal. To reduce this effect the ground current should be kept minimum and the ground impedance kept as low as possible. The existence of the ground currents may be determined by measuring the ground lead current with an ammeter or clip-on current probe. To reduce the ground current connect all the equipment in the setup to the ground at one point, use insulation transformers or connect a several hundred ohm resistor in series with the ground prong of the power cord. This can easily be accomplished by using a 3 to 2 prong cord adapter and connecting the resistor in series with the ground wire from the adapter.

The impedance of ground connection may be made smaller by using as short a ground lead as possible. Model 320A has a coaxial input which provides a low ground impedance. Use this coaxial input for measurement above 1 Mc wherever possible. For low frequency measurement the input may be converted to binding posts using Model 617 adapter.

2.5.6 Transmission Line Error is caused by transmission line effect in unmatched input leads. When 50 Ω cable is used this effect should be considered when measuring above 3 Mc and when line length exceeds 3 ft.

2.5.7 Magnetic and Electrostatic Fields may cause pickup in the input leads or induce disturbances into the sensitive parts of the voltmeter. To reduce this effect use coaxial, shielded twisted or closely spaced input leads. Do not place the voltmeter close to strong sources of magnetic or electromagnetic fields. The most sensitive part of the voltmeter is the lower right corner where the input circuitry is located.

2.6 Use of the DB Scale

The db scale on Model 320A is based on dbv (1 mW into 1000 Ω) reference. To determine the dbv level, add the db indication from the red db scale to the db figure indicated on Range Selector. The dbv level may be converted to other reference levels by using conversion figures shown in Table I.

TABLE I

DB LEVEL CONVERSION	
Reference Level	**Conversion from dbv (1 mW/1000 ohms)**
0.775 V, 1 mW in 600 Ω (dbm)	add 2.2 db
1.732 V, 6 mW in 500 Ω	subtract 4.8 db
1.897 V, 6 mW in 600 Ω	subtract 5.6 db

To simplify the conversion and to compute db levels of other references a handy db slide rule is available from Ballantine Laboratories on request, free of charge.

2.7 Effect of Harmonics

In case of complex waveforms with several harmonics the Model 320A measures the true rms voltage independent of the phase relationship of the harmonics.

$$E_{rms} = \sqrt{E_1^2 + E_2^3 + E_3^2 + \ldots + E_k^2} \quad (1)$$

E_{rms} — rms amplitude of the complete waveform

E_1 — rms amplitude of fundamental frequency component

E_2, E_3 — rms amplitude of harmonic frequency components

When some of the harmonic frequencies are beyond the voltmeter's frequency range or fall into less accurate range, their effect may be calculated by using the equation (1).

The voltmeter response, although not calibrated, extends above the specified 4 Mc limit. Except on the 0.3 mV range, the response is down approximately 1.5 db at 6 Mc and 3 db at 7 Mc.

The use of the equation (1) is illustrated with the following example:

Example:

A 500 kc square wave with the rms amplitude of 1.000 V is measured. To achieve an accuracy of 0.1% the harmonics up to 51 harmonics with maximum frequency of 25.5 Mc, should be measured. When the Model 320A is used for measurement, the bandwidth is limited to 7 Mc or to 13th harmonic. Under those conditions the result would be 0.989 V or an error of 1.1%.

To illustrate the effect of reduced accuracy for different harmonics, an assumption is made that the Model 320A used for measurement has the maximum negative error allowed by specification, as shown in Table II.

TABLE II

MAXIMUM CORRECTION FOR HARMONICS	
Harmonic	**Error**
1.5 Mc (3rd harm)	—3%
2.5 Mc (5th harm)	—3%
3.5 Mc (7th harm)	—5%
4.5 Mc (9th harm)	—5%
5.5 Mc (11th harm)	—10%
6.5 Mc (13th harm)	—30%

This assumption is not realistic because the voltmeter normally has positive and negative errors which tend to compensate one another. Using the formula for this hypothetical case, the result would be 0.984 V or an error of 1.6% or only 0.5% less than when measured with flat response instrument.

As a conclusion, the measuring accuracy is determined mostly by the accuracy of fundamental frequency or harmonics which have the highest energy contents. The harmonics with lesser energy contents have a small effect proportional to their energy contents (square of the amplitude). The effect may be calculated using equation (1) by considering only the harmonics which fall into the passband of the voltmeter and/or by assigning the harmonics amplitudes corrected by the known error of the voltmeter.

2.8 Effect of High Crest Factor

The Model 320A measures with specified accuracy any complex waveform with the maximum crest factor (ratio of peak and rms amplitude) — 15 for a down scale indication — decreasing proportionally to a maximum 5 for full scale indication.

The crest factor may exceed those limits without affecting measuring accuracy if the following conditions are fulfilled:

a) the peaks occur only occasionally
b) the maximum crest factor of the peaks does not exceed 8 (fs) to 25 (ds)
c) the peaks have sharp or rounded tops

Higher crest factor than 8 (fs) to 25 (ds) always causes high error because of amplifier overload.

2.8.1 Flat Top Waveforms (pulses) with crest factor of more than 4 (fs) to 10 (ds) cause negative errors,

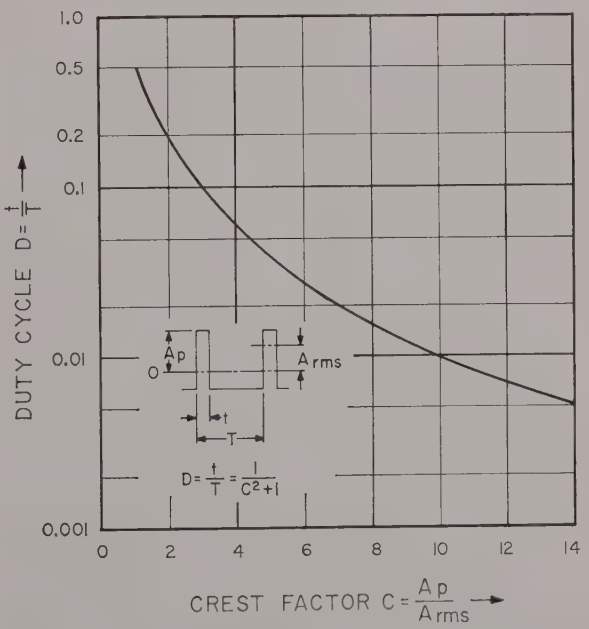


Fig. 4. Duty Cycle and Crest Factor of a Pulse Train

which may be as high as —2% for maximum crest factor 5 (fs) to 15 (ds). The error will increase rapidly with higher crest factor pulses, and the use of the voltmeter is not recommended.

The relationship of the crest factor to the duty cycles of a pulse train is shown in figure 4.

2.9 Effect of the DC Component

The Model 320A responds only to the ac component of a waveform. If an unsymmetrical waveform which contains a dc component has to be measured, measure the dc component separately with a dc voltmeter and compute the true rms value by the equation (2):

$$E_{rms} = \sqrt{E_{dc}^2 + E_{ac}^2} \qquad (2)$$

Example:

1 V dc is chopped into square pulses with 1 V peak amplitude. This pulse train has a dc component of 0.5 V which will be measured if the waveform is fed into a dc voltmeter. The ac component of the waveform is a 0.5 V amplitude square wave with rms value 0.5 V which will be measured when the Model 320A is used.

The total rms value of the waveform is

$$E_{rms} = \sqrt{E_{dc}^2 + E_{ac}^2} = \sqrt{0.5^2 + 0.5^2} = 0.707 \text{ V}$$

which will be measured with thermocouple-type instruments which respond both to ac and dc components.

The dc component of the waveforms is blocked by the input capacitor which has a 600 V dc rating. When more than 600 V dc is superimposed on the ac voltage, use an external 0.1 μF or higher blocking capacitor with proper dc rating and connect a 20 MΩ resistor to the input as shown in figure 5.

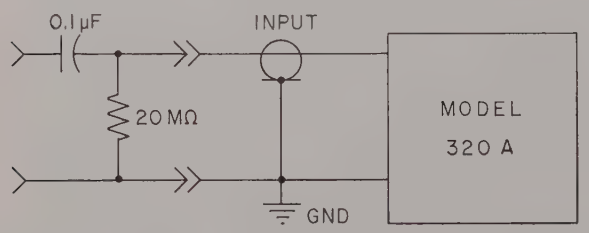


Fig. 5. Connection of the External Blocking Capacitor

2.10 Measurement of Noise and Random Signals

On noise and random signal measurement an error may be introduced by the limited bandwidth and by the limit of maximum crest factor. In case of white noise, the rms value of the noise voltage E_{rms} depends on the bandwidth:

$$E_{rms} = K \sqrt{\Delta f} \qquad (3)$$

Δf — noise bandwidth (cps)

K — factor depending on the source

If the bandwidth of noise source is wider than the bandwidth of Model 320A an error in noise measurement is introduced. This error may be calculated from the equation (3).

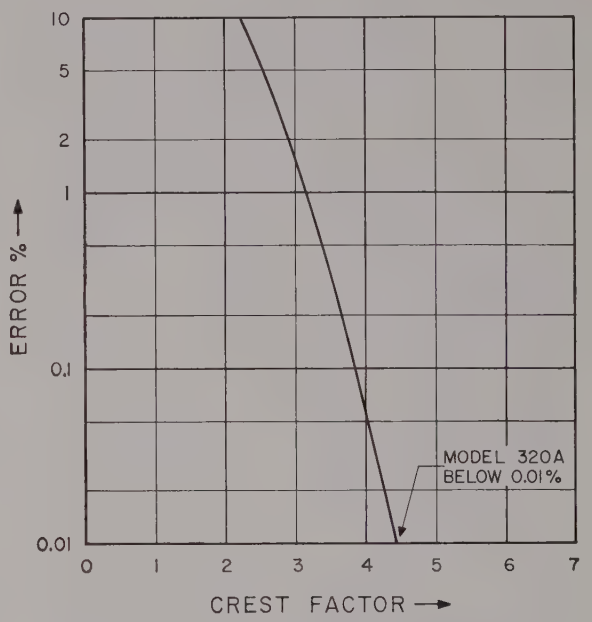


Fig. 6. Crest Factor Error with White Noise

For wide band noise measurement, 7 Mc may be considered as maximum bandwidth.

The crest factor limit also introduces an error. When measuring noise or random occurrences with Gaussian amplitude distribution, this error is depicted in figure 6. In the case of the Model 320A the crest factor limit of 5 introduces less than 0.01% error, which is negligible. However, when some circuit elements or preamplifiers introduce a more severe limitation of crest factor, the error should be considered when maximum crest factor limit approaches 3.5.

2.11 Measurement of Waveforms with Higher Crest Factor than 5

Waveforms up to crest factor 10 may be measured when the indication is limited to the lower half of the scale. To accomplish this, the input signal may be attenuated by a factor of 2 every time the indication comes to the upper half of the scale. An external attenuator, connected in front of the voltmeter, as shown on figure 7, may be used.

When the unknown voltage produces an indication on lower part of the scale, measure with switch in X1 position and read directly from scale. When indication falls in the upper part of the scale, set switch to X2 position and multiply the

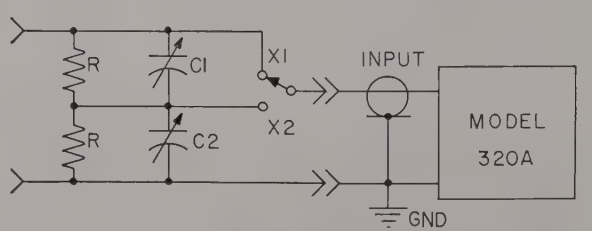


Fig. 7. Use of External Attenuator to Extend the Crest Factor Limit

indication by 2. Match the attenuator resistors to 0.1%. Select value depending on loading restriction, to the source. In case of high values (above 1000 Ω) and when measuring high frequencies, the attenuator should be compensated by capacitors C1 and C2. These should be adjusted to the accurate 2 to 1 attenuation ratios at high frequency.

A ratio transformer may be used for low frequency measurements.

2.12 Measurement on .3 mV Range

The effect of 24 μV internal noise level is negligible on all ranges except in 0.1 mV to 0.3 mV range. In this range it causes approximately +3% error for 0.1 mV signal and 0.3% error for 0.3 mV signal. Because of the noise and reduction of amplifier feedback, the accuracy of measurement in the 0.3 mV range is reduced. The effect of noise can be corrected, reducing the indication by the amounts shown in Table III.

TABLE III

CORRECTIONS ON 0.3 mV RANGE
3% for 0.09 mV to 0.11 mV indication
2% for 0.11 mV to 0.14 mV indication
1% for 0.14 mV to 0.2 mV indication
0.5% for 0.2 mV to 0.31 mV indication

Because of the reduced feedback, the falloff at the band ends is more rapid than on the other ranges. The limits for crest factor are the same as for all other ranges.

2.13 Measurement of Voltages Above 300 V

Voltages from 30 V to 10 kV may be measured with the Model 1320 plug-in multiplier. This multiplier has a multiplication ratio of 100 and has an input resistance of 40 MΩ in parallel with 3 pF. The multiplier is usable from 5 cps to 1 Mc and causes an additional measurement uncertainty of ±2%.

Voltages from 1 V to 10 kV may be measured with the Model 1311 High Voltage Probe. This probe has a multiplication ratio of 10,000 and an input resistance >10,000 MΩ in parallel with 4.5 pF. The probe and Model 320A are usable from 8 cps to 1 Mc with an additional measurement uncertainty of ±1%.

2.14 Use as Null Detector

In the null detector mode, the voltmeter may be used for approximate measurements and nulling of voltages normally below the scale.

2.14.1 To Use the Model 320A as Null Detector

Set	To
Function Switch	NULL DETECT
Range Selector	Any Range

Now the pointer's rest position is moved from against the left hand stop to the lower part of the scale so that small changes in deflection can be observed. The lowest observable voltage change in 0.3 mV range is 1 to 2 μV. Therefore a null may be observed with very high precision.

2.14.2 For Approximate Voltage Measurements Below 100 μV

Set	To
Function Switch	NULL DETECT
Range Selector	0.3 mV

Short the input, adjust the NULL ADJ on the rear until pointer indicates 0 db on red scale. Apply voltage to the input and observe the indication. Use figure 8 to convert the indication to input voltage.

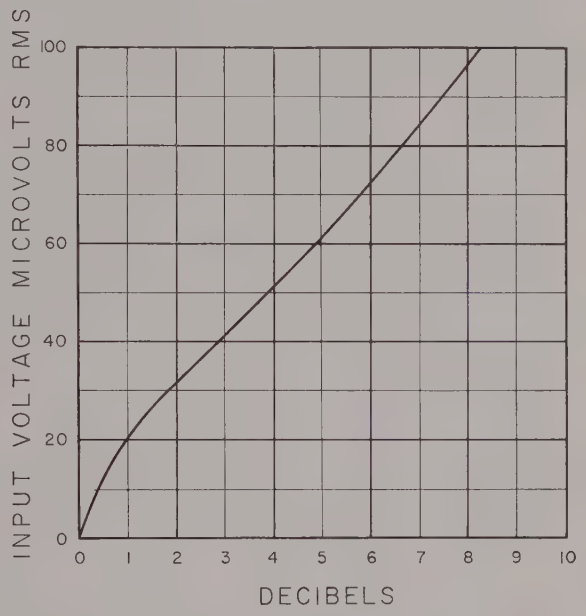


Fig. 8. Calibration of 0.3 mV Range on Null Detector Mode

The frequency response in Null Detector mode is the same as in RMS METER mode. The amplitude response deviates slightly from true rms. A small pointer jitter, caused by the noise of the voltmeter, is normal.

Return Function Switch to RMS METER position after finishing the measurement. This will avoid possible errors in subsequent use as voltmeter, when the position of Function Switch may be overlooked.

2.15 Overload Characteristics

The safe overload limits for extended periods are

Range	Max Voltage
0.3 to 10 mV	35 V
30 mV to 300 V	350 V

For a shorter period higher overloads up to 300 V on the most sensitive range will cause no damage.

2.16 Influence of Power Line Voltage

The Model 320A is operative over a 90 V to 130 V (180 V - 260 V) line voltage and 50 cps to 420 cps line frequency range. For ±10% variation of line voltage from 115 V (230 V) nominal voltage, the typical change in voltmeter sensitivity is given in Table IV.

TABLE IV

EFFECT OF LINE VOLTAGE		
Voltage Range	**Frequency Range**	
	20 cps - 1 Mc	**5 cps - 4 Mc**
1 mV to 300 V	±0.2% max	±0.5% max
0.3 mV	±0.3% max	±1.0% max

In null detector mode the pointer rest position is slightly affected by line voltage, readjust for accurate measurements.

2.17 RMS Current Measurements

The Model 320A may be used together with resistances of known value for making rms current measurements. The Ballantine Model 600 series of precision wirewound shunt resistors is recommended for this purpose. These resistors are available in the values of 0.01, 0.1, 1, 10, 100 and 1000 ohms and are supplied in plug-in form for insertion across the INPUT terminals. With these resistors, rms current of from 0.1 microamperes to 10 amperes can be measured. The crest factor limit is the same as for rms voltage measurement. Frequency range is dependent on the shunt used.

2.18 Power Dissipation Measurement

Because of the true rms response, the Model 320A may be used in power dissipation measurements in a pure resistive load. A voltage E (V) across a resistor R (Ω) dissipates

$$P = \frac{E^2}{R} (W) \qquad (4)$$

Across a 600 Ω resistor the Model 320A measures dissipation from 16.7×10^{-12} to 180 W. This corresponds to 10^{13} power ratio or 130 db. DB ratio could be read directly from the db scale of the voltmeter. To convert that to standard reference level, such as dbm or other, use constants listed on Table I.

2.19 Use of Mean Square Output

Model 320A may be used as a squaring ac to dc converter For this mode,

Set	To
Function Switch	MEAN SQUARE
Range Selector	According to the Input Voltage

On any range, the input voltage which produces a full scale deflection in voltmeter mode of operation produces a −0.2 V output. This changes proportional to the square of input voltage. The source resistance of output is 1000 Ω in parallel with 50 μF. The output follows the square law to a maximum instantaneous input amplitude of 5 times the full scale deflection. The maximum dc output is −1.0 V. With higher than 0.2 V output, the approximation of the square law has a negative error depending on the waveform.

A closer approximation of square law over a wider range of input voltage is possible when the R76 1 kΩ resistor is changed to lower value. This change will proportionally decrease the output voltage.

The dc output reaches 90% of its final value in 0.1 second. The ripple on dc output is negligible at 1 kc and 50% at 5 cps. Longer or shorter response or averaging times can be achieved by changing the value of C56 50 μF capacitor. This will also decrease or increase the ripple. Use tantalytic capacitors with minimum voltage rating of 6 V.

In case a low resistance recorder or external indicator is connected to MEAN SQUARE output, the load resistor R76 may be disconnected. The current output from MEAN SQUARE connecter is −200 μA ±10% corresponding to full scale indication. Maximum available current is −2 mA.

The external device connected to MEAN SQUARE output should not apply any dc voltage to the output. This will affect the accuracy of square law.

2.20 Use as Amplifier

The voltmeter may be used as a stable, wide band amplifier.

Set	To
Function Switch	AMPLIFIER
Range Selector	According to the Input Voltage or Desired Gain
Output	Binding Posts at Rear

The maximum gain is 90 db ±1 db bandwidth 5 cps to 4 Mc. Range Selector reduces the gain in 10 db steps until the lowest gain of −30 db on 300 V range.

The amplifier has a balanced transformer output which can be used floating, single-ended, or center tap grounded. For wide band operation ground the center tap.

The output from the output terminal marked IN PHASE is in phase with the input voltage. Voltage on REVERSE terminal has 180° phase shift with respect to input voltage.

The output voltage between IN PHASE and REVERSE terminals is 10 V when a signal which produces a full scale deflection in voltmeter mode is applied to the input.

The maximum distortion-free output signal is 50 V, for frequencies 30 cps to 500 kc. Outside this band the maximum available output voltage decreases to 10 V at 5 cps and 4 Mc. The voltage and gain between CT and IN PHASE or REVERSE terminals is 6 db less than the figures given above.

The source impedance at midfrequency is approximately 170 Ω from CT to IN PHASE or REVERSE terminal or 340 Ω from REVERSE to IN PHASE terminal. Between IN PHASE and REVERSE terminal the minimum allowable load resistance for 50 V maximum output is 20 kΩ, for 10 V maximum output 2 kΩ. The minimum loading resistor across half of the transformer output can be one-half of that value. Capacitive loading should be limited to 25 pF.

A maximum +300 V or −100 V dc can be applied between output transformer winding and chassis without causing damage.

3. CIRCUIT DESCRIPTION

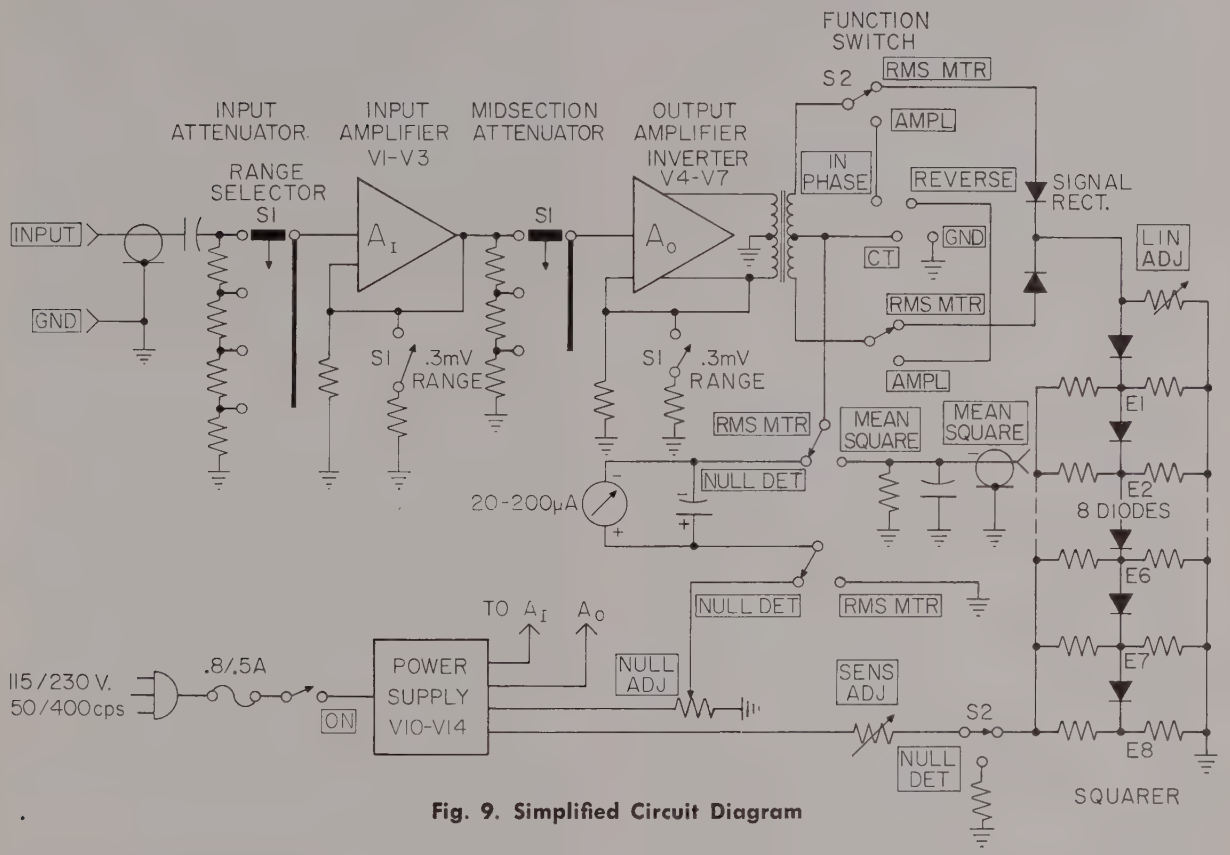


Fig. 9. Simplified Circuit Diagram

The basic voltmeter circuits are shown on simplified diagram figure 9. A complete schematic is at the end of the instruction book.

The basic parts of the voltmeter are:

3.1 Input and Midsection Attenuators

Two wide band attenuators are used in the Model 320A — the input attenuator at the input and the midsection attenuator between the amplifiers.

The voltmeter sensitivity change on all ranges is achieved by input and midsection attenuator except on 0.3 mV range where the gain of the amplifiers is increased.

3.1.1 Input Attenuator is connected to the voltmeter input over dc blocking capacitor. It consists of two separate attenuators R4, R6, and R4, R7, R8.

Below 20 kc the attenuator acts as a resistive divider; above 20 kc the attenuator is capacitive and attenuation ratio is adjustable with C6, C7 and C6, C8, respectively Stable, closely matched (0.1%) hermetically-sealed, deposited film resistors are used in attenuator. The attenuation ratios and the scale ranges on which the attenuator is in the circuit are shown in Table V.

TABLE V

ATTENUATION RATIOS

Voltage Range FULL SCALE	Input Attenuation Ratio	Midsection Attenuation Ratio	Total Attenuation Ratio	Total Gain of Amplifiers
.3 mV	1	1	1	84 db
1 mV	1	1	1	74 db
3 mV	1	3.162	3.162	74 db
10 mV	1	10	10	74 db
30 mV	31.62	1	31.62	74 db
100 mV	31.62	3.162	100	74 db
300 mV	31.62	10	316.2	74 db
1 V	1000	1	1000	74 db
3 V	1000	3.162	3162	74 db
10 V	1000	10	10000	74 db
30 V	31620	1	31620	74 db
100 V	31620	3.162	100000	74 db
300 V	31620	10	316200	74 db

3.1.2 Midsection Attenuator is a string type attenuator with resistors R40, R41 and R42. Because of the lower resistance values it acts as a resistive attenuator

over most of the voltmeter frequency range. Slight compensation at higher frequencies is achieved with C30 and C31.

The attenuation ratios with corresponding ranges are shown in Table V.

3.2 Input Amplifier is a stable, three-stage (tubes V1, V2 and V3) feedback amplifier with a gain of 135 and with approximately 42 db of overall feedback. On 0.3 mV range the gain of the amplifier is increased to 240 by shunting the feedback resistor R20. R29 serves to vary the gain slightly on this range.

The high frequency response of the amplifier is adjustable slightly with L1 on all ranges and with C20 on 0.3 mV range only.

3.3 Output Amplifier is also a stable, three-stage (V4, V5, ½ V7) feedback amplifier with a gain of 38 and with approximately 38 db of overall feedback. The last stage of the amplifier is a direct coupled cathode follower for low output impedance. By shunting part of the feedback resistor the amplifier gain is increased to 67 on 0.3 mV range. A slight high frequency adjustment on all ranges is possible with C36 and C44.

It drives one-half of the output transformer. The other half is driven by the phase inverter.

3.4 Phase Inverter is a two-stage (V6, ½ V7) phase reversing amplifier with the gain of 1 and with approximately 37 db of overall feedback. Together with output amplifier it drives the push-pull output transformer. The symmetry and balance of the drive is controlled by matched resistors R65 and R66 at low frequencies and with C46 at high frequencies.

3.5 Square Law Detector

The true rms response is achieved in the square law detector where the square law is synthesized by continuous sequence of linear segments. With this technique a wide band, instantaneous responding, wide dynamic range, overload-proof and accurate rms response is accomplished. The ideal characteristic of a square law detector is

$$I_s = K \bullet E_s^2$$

I_s — squarer current

E_s — voltage applied to the squarer

K — constant

The detector in the Model 320A approximates this law to $\pm 0.3\%$ for smoothly varying waveforms and to $\pm 0.5\%$ for flat top waveforms.

The squarer consists of 8 series connected diodes CR4 - CR11. At every junction of diodes a pair of resistors (R82 - R91 for diode junction CR4 - CR5, etc.) supplies a dc bias and load to the junction. The bias voltage is increasing and the load resistance decreasing from diode CR4 to diode CR11. The signal, which is rectified in CR1, CR2 full wave rectifier passes through only the diodes whose bias is less than the instantaneous amplitude of the signal. At higher signal level more diodes are conducting, less load resistance is applied to the signal, and the squarer draws a current proportional to the square of the signal voltage. The rectified and squared current passes through a logarithmic indicator with 10 - 1 current range and 200 μA full scale sensitivity. Because of the square law of the detector, this current variation is produced by 3.16 - 1 input voltage range. The logarithmic scale of the indicator is retained with a voltage range of 10 db, even though the indicator has a current range of 20 db.

Capacitor C55, together with indicator movement inertia, averages the squared current. The diode CR3 is to protect the indicator from overload.

The sensitivity of the voltmeter is changed by changing the sensitivity of the square law detector by controlling the bias of all diode junctions with R71 SENS ADJ control. The temperature affects the forward characteristics of diodes, which are then compensated by the thermistor RT1.

In MEAN SQUARE output mode a 1000 ohm resistor R76, with averaging capacitor C56, is connected into the squarer circuit instead of the meter.

In NULL DET mode the squarer sensitivity is increased by disconnecting the bias on all diodes, and applying a small current in series with rectifiers and squaring network, which can be controlled by R77 NULL ADJ.

In AMP OUTPUT mode the squarer is disconnected and the transformer output connected to the output terminals for an ac output linearly proportional to the input voltage.

3.6 Power Supply

Two dc power supplies are used, one with silicon rectifiers CR12 and CR13 to supply dc filament power to the tubes V1 and V2. A regulated high voltage supply with V10 as rectifier supplies the high voltage to amplifiers and squarer. The B+ supply to the amplifiers is stabilized by the V11 and V12 cold cathode regulator tubes, and to the output cathode follower by V13 series regulator tube. A separate reference tube V14 stabilizes the bias voltage for squarer.

4. MAINTENANCE

4.1 General

The Model 320A is designed for extended trouble-free service. The use of high grade components operated below rating and low operating temperature results in more than 5,000 hours expected operating periods without the need for amplifier tube change.

The purpose of periodic maintenance procedures is to assure the specified accuracy, check the condition of the voltmeter and detect and replace deteriorated tubes or components before they have an appreciable effect on instrument performance. The procedure is designed not only to measure and correct voltmeter performance at the time of test, but to foresee the possible troubles and guarantee the instrument accuracy through the next maintenance period.

TABLE VII

SQUARER BIAS LIMITS				
Diode Junction	**Voltage at Diode Junction*** Voltage on Terminal 43			**Limit**
	65	70	75	
CR1, CR2, CR4	0.02	0.02	0.02	.03 max
CR4, CR5	0.50	0.54	0.58	±3%
CR5, CR6	0.84	0.90	0.97	±3%
CR6, CR7	1.25	1.35	1.45	±3%
CR7, CR8	1.87	2.0	2.15	±3%
CR8, CR9	2.8	3.0	3.2	±5%
CR9, CR10	4.2	4.5	4.8	±5%
CR10, CR11	6.3	6.8	7.3	±5%
R89, R98	10.4	11.1	11.8	

*Interpolate the voltage values if the terminal voltage is not exactly 65, 70 or 75 V.

5.6.4 Electrical Linearity (R90) needs adjustment when the voltmeter has normal full scale sensitivity but the meter screw has inadequate range for bottom scale adjustment. The electrical linearity control R90 is located at the squarer board and may be adjusted through the hole marked LIN ADJ at the rear of the case.

5.6.4.1 For the adjustment turn the instrument on, allow 20 minutes warmup, then

Set	To
Function Switch	MEAN SQUARE
Range Selector	1 V

Apply a 1 V signal to the input. Connect a sensitive dc voltmeter to the MEAN SQUARE output connector on the front panel of the Model 320A. It should indicate approximately —0.2 V. Readjust the input signal until the output is —0.2 V ±0.5%.

5.6.4.2 Turn Range Selector to next higher (3 V) range. Adjust the R90 LIN ADJ until the output is —0.0206 V ±0.5%. Repeat the steps 5.6.4.1 and 5.6.4.2 until the ouput is in specified limits at either range selector setting.

5.7 Frequency Response Alignment

Typical frequency response of the voltmeter with 1 kc response as reference is shown on figure 11.

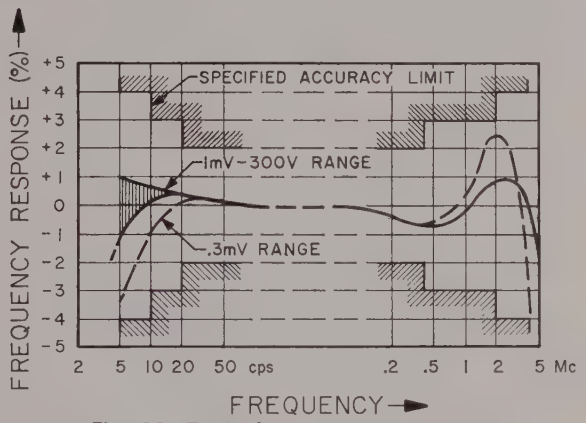


Fig. 11. Typical Frequency Response

If the frequency response at any range differs more than ±1 to 2% from the typical response shown in the graph, the response alignment is needed.

5.7.1 Low Frequency Response has no alignment controls. Its deviation from the normal response is

TABLE VIII

RESPONSE ALIGNMENT CONTROLS				
Order of Alignment	**Assembly**	**Adjustment**	**Affects Ranges**	**Adjust on Range**
1	Amplifier	C44	all ranges	1 mV
2	Midsection Attenuator	C30	3, 100 mV; 3, 100 V	3 mV
3	Midsection Attenuator	C31	10, 300 mV; 10, 300 V	10 mV
4	Input Attenuator	C6	30 mV and up	1 V
5	Input Attenuator	C7	30, 100, 300 mV	30 mV
9	Input Attenuator	C8	30, 100, 300 V	30 V
7	Amplifier	C20	0.3 mV	0.3 mV

In case of an error on several ranges, this table helps to determine the control which has to be adjusted. A more detailed alignment procedure is outlined in Table IX at the rear of this instruction book.

caused by the changes of amplifier coupling or decoupling networks or deterioration of amplifier tubes. The most probable cause is the development of grid current in some stage. Normal response will be restored after the replacement of defective component or tube.

5.7.2 High Frequency Response may deviate from normal response because of the change in amplifier response or changes in attenuator compensation.

Amplifier response adjustment affects the responses of all ranges.

The attenuator adjustment offsets only certain ranges as indicated in Table VIII.

5.7.3 Amplifier Response Alignment is needed when the response of all ranges deviates from normal response.

Two controls are available for alignment. The Control C44 aligns the ranges from 1 mV to 300 V and a separate control C20 aligns the 0.3 mV range. C44 adjustment affects also 0.3 mV range and should be done before this range is adjusted.

5.7.3.1 For amplifier response adjustment remove the case, allow 5 minutes warmup and connect stable 1 kc to 5 Mc signal source 0.5 - 1 mV output to the input (Ballantine Model 440 with 5 mA thermocouple and 0.22 Ω resistor) as shown on figure 10.

Two steps are necessary for amplifier response alignment.

5.7.3.2. Phase Inverter Balance Check. This is needed when resistor R65, R66, R67 or tube V6 has deteriorated.

Set	To
Function Switch	AMPLIFIER
Range Selector	1 mV

Apply a 0.5 to 1 mV 1 kc signal to the input. Measure the voltage from CT to IN PHASE and REVERSE output terminal using diode voltmeter (5.2.7). The voltage at both output terminals should be equal within ±0.2%. If the voltage differs more, check R65, R66, and R67. R65 and R66 are matched pair to 0.1%.

5.7.3.3 Change the input frequency to 3 Mc. Measure the voltage at the same output terminals. If the voltage differs more than 0.2% adjust the voltage on IN PHASE terminals with C46 HF BAL control which is located at the tube side of output amplifier printed board until the voltage agrees with voltage at REVERSE terminal.

5.7.3.4 Amplifier Response Alignment. For the adjustment remove the case and

Set	To
Function Switch	RMS METER
Range Selector	1 mV

Apply a 1 kc 0.5 - 1 mV signal to the voltmeter through Model 440 Micropotentiometer (5 mA thermocouple with 0.22 Ω resistor). Adjust the signal to produce an indication directly to some scale line on the indicator. Note micropotentiommeter dc output.

5.7.3.5 Change the signal frequency to 3 Mc. Adjust the RF amplitude exactly to the 1 kc amplitude applied to the voltmeter before by adjusting the level to the same dc output from micropotentiometer.

The indication on voltmeter should be ±0.5 to +1.5% above the previous indication. If not, adjust C44 on the output amplifier board until the indication is ±0.5 to +1% above the 1 kc indication.

5.7.4 0.3 mV Range Adjustment is needed when the response of 1 mV range is normal but 0.3 mV differs from the normal response. For the adjustment remove the case and

Set	To
Function Switch	RMS METER
Range Selector	.3 mV

Apply a 0.3 mV 1 kc signal to the voltmeter input through Model 440 Micropotentiometer (5 mA thermocouple with 0.22 Ω resistor). Adjust signal to produce indication exactly on some scale line; note the micropotentiometer dc output.

5.7.4.1 Change signal frequency to 3 Mc. Adjust rf amplitude exactly to 1 kc amplitude by adjusting the level to give the same dc output from the micropotentiometer. Adjust C20 through the left hole (see figure 13) on the shield at input amplifier board until the 3 Mc indication is within +2 to +3% of 1 kc indication.

5.7.5 C30 Midsection Attenuator Adjustment is needed when the voltmeter response on 1 mV range is normal but deviates from normal on 3, 100 mV and 3, 100 V range. For the adjustment, remove the case and

Set	To
Function Switch	RMS METER
Range Selector	3 mV

Apply a stable 2 - 3 mV 1 kc signal to the input. Use Model 440 Micropotentiometer (5mA thermocouple with 2.2 Ω resistor). Adjust signal to produce an indication at some scale line between 2 and 3. Note the input level.

5.7.5.1 Change signal frequency to 3 Mc. Adjust the amplitude exactly to 1 kc amplitude by adjusting level to give same dc output from the Micropotentiometer. Adjust C30 until voltmeter indicates +1 to 2% above the 1 kc indication. C30 is the top trimmer on Range Selector rear deck as shown in figure 14.

5.7.6 C31 Midsection Attenuator Adjustment is needed when voltmeter response is normal on 1 mV range but deviates from normal to 10, 300 mV and 10, 300 V ranges. For the adjustment, remove the case.

Set	To
Function Switch	RMS METER
Range Selector	10 mV

Apply a stable 1 kc 5 mV - 10 mV signal to the voltmeter, through Model 440 Micropotentiometer (5 mA thermocouple with 2.2 Ω resistor). Adjust the signal for an indication on some scale line. Note the micropotentiometer dc output.

5.7.6.1 Change the signal frequency to 3 Mc. Adjust the amplitude exactly to 1 kc amplitude by adjusting level to give same dc output from the Micropotentiometer. Adjust C31 until voltmeter indicates 0 to 1% above 1 kc indication. The C31 is the lower trimmer on Range Selector near deck as shown in figure 14.

5.7.7 Input Attenuator Adjustments are made with the voltmeter in case. The trimmers are accessible through the front panel openings after removal of the escutcheon plate. For removal of escutcheon plate remove the function switch and range selector knob. Remove four #4 nickel plated binder head screws at the corners of the escutcheon plate and lift off the escutcheon plate. The adjustment holes under the escutcheon plate are marked on the panel.

Adjust the input attenuator in sequence shown on Table VIII or IX.

5.7.7.1 C6 Input Attenuator Adjustment is needed when voltmeter response on 1 mV range is normal but differs from normal on ranges from 30 mV and up.
For the adjustment:

Set	To
Function Switch	RMS METER
Range Selector	1 V

Apply a stable 1 V 1 kc signal to the voltmeter. Monitor the signal with Model 393 Transfer Voltmeter (with 1393-2 probe) as shown in figure 11. Adjust the signal to produce indication exactly on some scale line. Register the level by balancing AC BALANCE on Model 393.

5.7.7.1.1 Change signal frequency to 50 kc. Adjust the level exactly to 1 kc level using Model 393 transfer voltmeter. Adjust C6 using insulated screwdriver until voltmeter indication agrees with 1 kc indication to ±0.2%.

5.7.7.2 C7 Input Attenuator Adjustment is needed when voltmeter response on 1 mV range is normal but differs from normal on 30 mV, 100 mV and 300 mV range.

For the adjustment:

Set	To
Function Switch	RMS METER
Range Selector	30 mV

Apply a stable 1 kc 20 - 30 mV signal to the voltmeter through Model 440 Micropotentiometer (5 mA thermocouple 6.8 Ω resistor). Adjust the signal to produce indication exactly on some scale mark. Note the dc output from Model 440.

5.7.7.2.1 Change signal frequency to 50 kc. Adjust signal amplitude exactly to 1 kc amplitude by adjusting the level to give same dc output from the Micropotentiometer. Adjust C7 until voltmeter indicator agrees with 1 kc indication to ±0.2%.

5.7.7.3 C8 Input Attenuator Adjustment is needed when voltmeter response on 1 mV range is normal but differs from normal on 30 V, 100 V and 300 V range.

For the adjustment:

Set	To
Function Switch	RMS METER
Range Selector	30 V

Apply a stable 30 V 1 kc signal to the input voltmeter. Monitor the signal with Model 393 Transfer Voltmeter (with 393-50 probe) as shown on figure 12. Adjust the signal to produce indication exactly on some scale line. Register the level by balancing AC BALANCE on Model 393.

5.7.7.3.1 Change signal frequency to 3 Mc. Adjust the level exactly to 1 kc level using Model 393 Transfer Voltmeter. Adjust C8 until voltmeter indication is +0.5 to 1.0% higher than 1 kc indication.

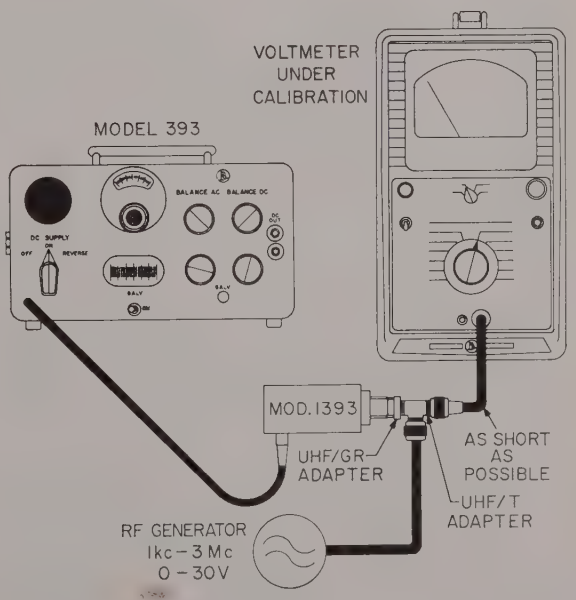


Fig. 12. Connection of HF Transfer Voltmeter

5.8 Sensitivity Adjustment

The voltmeter has two sensitivity controls. R71 changes the sensitivity on all voltmeter ranges and R29 affects the sensitivity of 0.3 V range only.

5.8.1 R71 Sensitivity Adjustment is needed when voltmeter exhibits similar error on all ranges at midfrequency (1 kc). Approximate adjustment may be accomplished at any range, 1 mV or above, by adjusting the R71 sensitivity control at the rear of the voltmeter until voltmeter indication agrees with the 0.4-2 kc calibration signal. Make the adjustment after 0.5 hour minimum warmup. Change defective tubes or make adjustments of the frequency response before adjusting sensitivity.

5.8.2 R29 .3 mV Range Sensitivity Adjustment is needed when only .3 mV range is in error at midfrequencies. If R71 adjustment is needed, it should be made before adjusting R29.

For the adjustment, remove the instrument from the case, allow 15 minutes warmup, and

Set	To
Function Switch	RMS METER
Range Selector	.3 mV

Apply a stable 1 kc signal 0.3 mV ±0.25% to the voltmeter and adjust R29 until voltmeter indicates 0.3 mV ±0.2%. R29 is located at the input amplifier board and is adjustable through the hole in the shield as shown in figure 13.

5.9 Table of Calibration Controls, their locations, ranges they affect and frequencies and limits of adjustments are shown at the end of the instruction book, Table IX.

5.10 Service of Attenuators

Stable, carefully matched resistors are used in the attenuator. They do not dissipate any appreciable power and therefore their failure is very unlikely. Before replacing any resistors investigate possible trouble sources such as grid current in tube V1 or V4, bad switch contact, etc.

For the replacement, check voltmeter error at midfrequency on ranges 3, 10, 30 mV and 1, 10 and 30 V. Tabulate the error. The error spread — maximum positive error less maximum negative error — should not exceed 0.3% because the attenuator resistors are matched to 0.1%.

If one of the ranges shows more error than allowed within the 0.3% spread, one or more of the midsection or input attenuator resistors must be defective. The defective string may be determined from the schematic or from the Troubleshooting Chart (Section 5.13).

5.10.1 To replace an attenuator resistor measure the dc resistance of all resistors in the attenuator string with 0.05% or better accuracy. The resistance value may vary ±1% from the nominal value indicated on schematic but all resistors should be matched to ±0.1%. Replace the resistor which differs more than 0.1% from correct matching. Use resistors of same type and make for replacement because different makes may have different high frequency characteristics.

5.10.2 After replacing the resistor, check error in all ranges and make sure that error spread does not exceed 0.3%. Adjust voltmeter sensitivity (R71 adjustment, section 5.8.1) so that maximum positive error equals approximately the maximum negative error. This condition gives the least measuring error.

5.10.3 Because of the close matching of the resistors a single replacement may not be obtained from Ballantine Laboratories and in most cases a whole string has to be replaced. For that reason it is recommended that the voltmeter be sent to Ballantine Laboratories for attenuator service.

5.11 Intrinsic Noise and Hum Test should be done after tube replacement (see section 4.7.4.1) or when instrument performs erratically on lower ranges, reads high on most sensitive range, or has high error when line frequency voltage is measured.

5.11.1 For noise check short the INPUT terminals of Model 320A.

Set	To
Function Switch	AMPLIFIER
Range Selector	.3 mV

Connect a wide-band voltmeter (4 Mc or more) to AMPLIFIER output terminals at the rear of Model 320A (high terminal to IN PHASE or REVERSE and low terminal to CT).

Measure noise voltage. Limit: 400 mV RMS maximum on .3 mV range and 125 mV RMS maximum on 1 mV range.

If noise is erratic or considerably higher, try new tubes starting in the order V1, V2, V4. VR tubes V11 and V12 may also be a source of erratic fluctuating noise.

The other possible noise sources are described in Troubleshooting Chart.

5.11.2 For hum check connect an oscilloscope to AMPLIFIER terminals of Model 320A.

Set	To
Function Switch	AMPLIFIER
Range Selector	.3 mV

Short INPUT terminals of Model 320A. Normal noise consists of wide-band noise with harmonics of power frequency relatively much lower or hardly noticeable. If high hum is present look for possible sources in Troubleshooting Chart.

5.12 Service of the Power Supply

The power supply dc voltages and typical ripple voltages are shown in the schematic at the end of the instruction book. In case of trouble, the dc and ripple voltage check will discover the most common faults.

The condition of VR tubes (V11, V12, V14) can be checked by lowering the line voltage to 100 V and checking the starting of the tubes at low line voltage. Sometimes excessive amplifier noise may be caused by the VR tubes which level may fluctuate slightly. Those tubes should be replaced.

The series regulator tube V13 for the output stage should have a bias of 5 V minimum between pin 1 and 2.7 with 115 V line voltage. Replace the tube with smaller bias.

5.13 Troubleshooting Chart

In the following chart the effects of possible malfunction of a component or maladjustments of controls are described:

Symptoms	Possible Cause	Pertinent Sections
Instrument inoperative Pilot light does not light up	No power; blown fuse; Defective power supply	5.3.3 5.12
Meter indicating with shorted input	Voltmeter in NULL DET mode; Defective tubes; defective coupling or decoupling capacitor or range selector switch, etc.	2.14 2.5.2
Meter indicating with no signal	Input not properly shielded; Voltmeter in NULL DET mode	2.5 2.14
Sensitive to mechanical shock	Microphonic tubes V1, V2, V4, etc.; Loose connection	5.11 5.5
High error at 60 cps measurement	Hum pickup through input wiring; High ground current; defective tubes V3, V4	2.5 5.11.2
High error at 120 cps measurement	Heater cathode short of V1, V2; High ripple of +B supply; open filter capacitors; inoperative VR tubes V11, V12; High dc heater ripple; Open bypass or decoupling capacitor	5.11 5.12
High error on lower part of the scale	Meter screw out of adjustment; Squarer control R90 out of adjustment; Leaky capacitors C47, C48 or output transformer; Leaky function switch wafer	4.5 5.6.4
Instrument is out of calibration: a. on all ranges at low frequencies	Voltmeter in NULL DET mode; R71 SENS ADJ off; Defective tube or component; Meter screw off; defective diode in squaring network	2.14 4.6.5 4.5 5.6.2, 5.6.3
b. on all ranges at frequencies above 1 Mc	Amplifier frequency response off; Deteriorated tube; defective component	5.7.3
c. on every third range at low frequencies	Defective resistor in midsection attenuator	5.10
d. on every third range at frequencies above 1 Mc	C30 or C31 misadjusted	5.7.5, 5.7.6
e. on three sequential ranges at low frequencies	Defective resistor in input attenuator	5.10

Troubleshooting Chart (Continued)		
f. on three sequential ranges at high frequencies above 30 kc	C6, C7 or C8 misadjusted	5.7.7
Sequential ranges don't agree	Meter screw or R90 LIN ADJ misadjusted; Diode CR3 leaky	4.5 5.6.4 5.6.3
Excessive flutter of pointer below 10 cps	Open C55	3.5
Pointer sluggish and positive errors at the bottom scale	Defective C55	3.5
High scale linearity error	Defective squarer diode or resistor	5.6

6. SHIPPING INSTRUCTIONS

If it should be necessary to return the instrument to Ballantine Laboratories, Inc. make certain that at least four inches of padding material surrounds the instrument to prevent damage during shipment. Ship via REA Express, motor truck, or air freight, to

Ballantine Laboratories, Inc.
90 Fanny Road
Boonton, New Jersey

TABLE IX

CALIBRATION TABLE

Order of Adjustment	CONTROL					
	Designation	Location	Controls	Affects Ranges	Adjust On Range	
1	R65, 66, 67	Output Amplifier Board TB3	L.F. Balance of Phase Inverter	All	1 mV	M
2	C46	Output Amplifier Board TB3	H.F. Balance of Phase Inverter	All	1 mV	M
3	C44	Output Amplifier Board TB3	Amplifier H.F. Response	All	1 mV	M
4	C20	Input Amplifier Board TB2	.3 mV Range H.F. Response	.3 mV	.3 mV	M
5	C30	Range Selector Rear Deck	Midsection Attenuator H.F. Response	3, 100 mV 3, 100 V	3 mV	M
6	C31	Range Selector Rear Deck	Midsection Attenuator H.F. Response	10, 300 mV 10, 300 V	10 mV	M
7	C6	Behind Escutcheon Plate	Input Attenuator H.F. Response	30 mV - 300 V	1 V	M
8	C7	Behind Escutcheon Plate	Input Attenuator H.F. Response	30, 100, 300 mV	30 mV	M
9	C8	Behind Escutcheon Plate	Input Attenuator H.F. Response	30, 100, 300 V	30 V	M
10	R71	Rear, Through Hole	Voltmeter Sensitivity	All	1 mV - 300 V	M
11	R29	Input Amplifier Board TB2	.3 mV Range Sensitivity	0.3 mV	0.3 mV	

CALIBRATION SIGNAL			Adjustment Limits %	Remarks
Recommended Source	Frequency	Accuracy		
ɔdel 420 Calibrator	1 kc	Not Important	±0.2%	Adjust for Equal Output from In Phase and Reverse Terminals to Chassis
ɔdel 440, 5 mA TC, 0.22 Ω Res	3 Mc	Not Important	±0.2%	
ɔdel 440, 5 mA TC, 0.22 Ω Res	3 Mc	±0.2% Reference 1 kc	+0.5 to 1.5%	Adjust in Reference of 1 kc
ɔdel 440, 5 mA TC, 0.22 Ω Res	3 Mc	±0.2% Reference 1 kc	+1 to 3%	Adjust in Reference of 1 kc
ɔdel 440, 5 mA TC, 2.2 Ω Res	3 Mc	±0.2% Reference 1 kc	+0.5 to 1%	Adjust in Reference of 1 kc
ɔdel 440, 5 mA TC, 2.2 Ω Res	3 Mc	±0.2% Reference 1 kc	+0.5 to 1%	Adjust in Reference of 1 kc
ɔdel 393 with 1393-2 Probe	50 kc	±0.2% Reference 1 kc	±0.1%	Adjust in Reference of 1 kc
ɔdel 440, 5 mA TC, 6.8 Ω Res	50 kc	±0.2% Reference 1 kc	±0.1%	Adjust in Reference of 1 kc
ɔdel 393 with 1393-50 Probe	3 Mc	±0.2% Reference 1 kc	0 to 1.5%	Adjust in Reference of 1 kc
ɔdel 420 Calibrator	1 kc	±0.2%	±0.1%	After Complete Warmup
	1 kc	±0.2%	±0.2%	After Complete Warmup

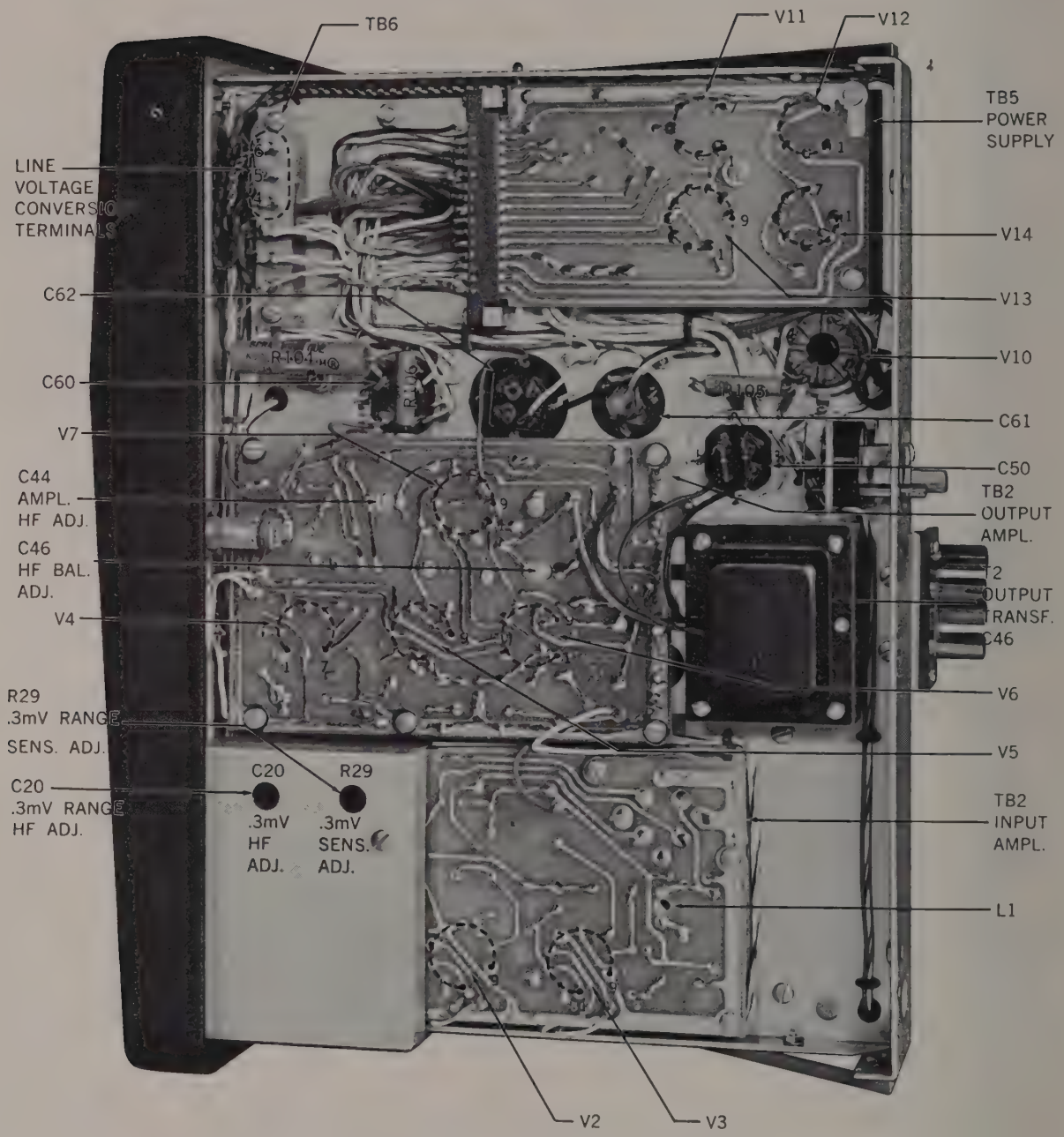


Fig. 13. Side View of Model 320A (right)
Showing Location of Tubes and Adjustments

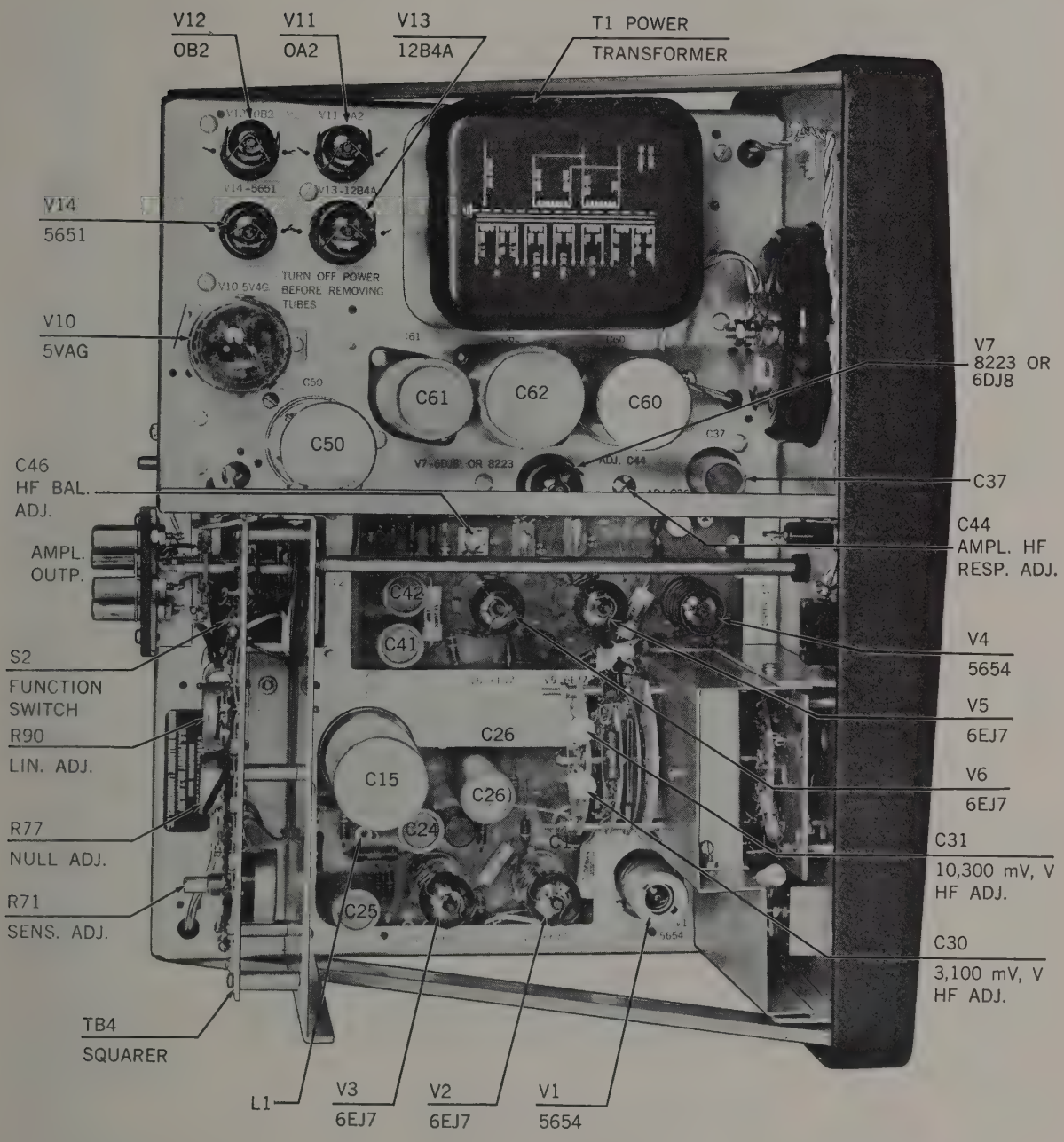


Fig. 14. Side View of Model 320A (left) Showing Location of Tubes and Adjustments

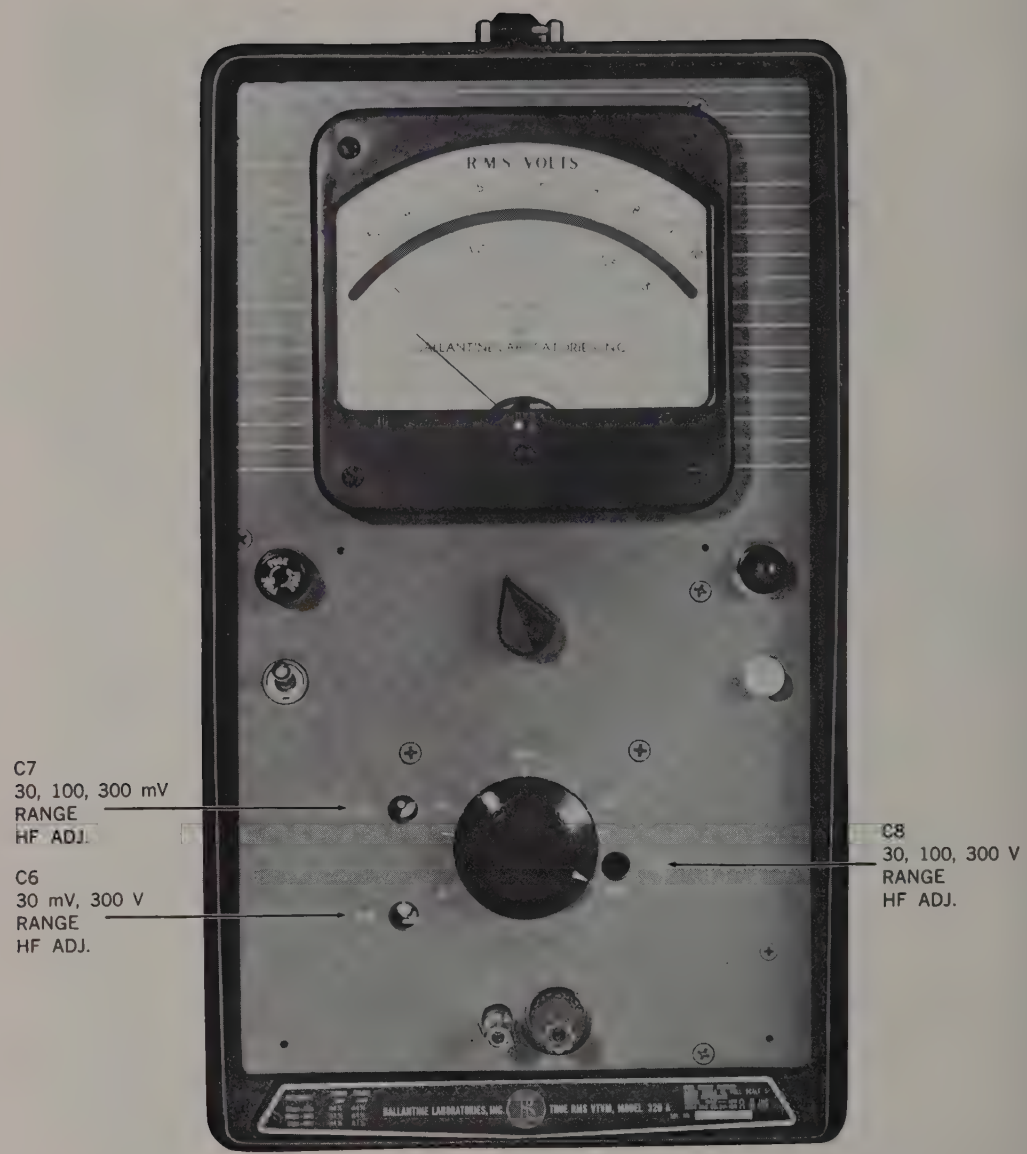


Fig. 15. Front View of Model 320A (escutcheon plate removed) Showing Location of Adjustments

7. REPLACEMENT PARTS LIST

REFER TO MODEL 320A CIRCUIT DIAGRAM, DWG. MD-1827L

B. L. Part No.	*Circuit Symbol*	*Capacitors*	*Manufacturer*
2388	C1	.033 μF, 10%, 600 V	Sprague
2255	C2	3300 pF, 5%, Type CM-20-E-332-J	Arco
2275	C3	75 pF, 1%, Type CM-15-E-750-E	Arco
2546	C4	.22 pF, ±10%, Type QC, 500 VDCW	Quality Components
2241	C5	15 pF, 5%, Type CM-15-C-150-J	Arco
2440	C6	1.2 - 3.5 pF, Type 189-1-5, 850 V	Johnson
2444	C7	2.5 - 11 pF, Type N300, Style 538-002	Erie
2441	C8	9 - 35 pF, Type N650, Style 538-002, 200 VDCW	Erie
2771	C15A	60 μF, −10%, +100%, 250 VDCW, Special	Ballantine
	C15B	40 μF, −10%, +100%, 250 VDCW, Special	Ballantine
	C15C	30 μF, −10%, +100%, 250 VDCW, Special	Ballantine
	C15D	30 μF, −10%, +100%, 250 VDCW, Special	Ballantine
2572	C16	4.7 μF, +20%, −10%, Type 150D475X0006A2, 6 VDCW	Sprague
2362	C17	.22 μF, 5%, Type 148P22492, 200 V	Sprague
2232	C18	150 pF, 5%, Type CM-15-E-151-J	Arco
2098	C19	12 μF, 250 V, Special	Ballantine
2412	C20	25 - 280 pF, Special	Ballantine
2364	C21	.033 μF, 5%, Type 148P33392, 200 V	Sprague
2535	C22	.01 μF, 0 +100%, Type BYA-6S1, 600 VDCW	Cornell-Dubilier
2094	C23	250 μF, −10%, +100%, Type 89 D 117, 3 VDCW	Sprague
2098	C24	12 μF, 250 V, Special	Ballantine
2098	C25	12 μF, 250 V, Special	Ballantine
2097	C26	50 μF, −10%, +100%, 250 VDCW	Ballantine
2442	C30	2 - 8 pF, Type NPO, Style 538-002, 200 VDCW	Erie
2442	C31	2 - 8 pF, Type NPO, Style 538-002, 200 VDCW	Erie
2371	C35	.047 μF, 10%, Type 148P47392, 200 V	Sprague
2277	C36	47 pF, 5%, Type CM-15-E-470J, 500 V	Arco
2098	C37	12 μF, 250 V, Special	Ballantine
2354	C38	.15 μF, 10%, Type 148P15492, 200 V	Sprague
2095	C39A	100 μF, −10%, +100%, Type 89D150, 12 VDCW	Sprague
	C39B	100 μF, −10%, +100%, Type 89D150, 12 VDCW	Sprague
2354	C40	.15 μF, 10%, Type 148P15492, 200 V	Sprague
2098	C41	12 μF, 250 V, Special	Ballantine
2098	C42	12 μF, 250 V, Special	Ballantine
2540	C43	.005 μF, 20%, Type EGP Z5U, Style 851, 500 VDCW	Erie
2440	C44	1.2 - 3.5 pF, Type 189-1-5, 850 V	Johnson
2545	C45	.82 pF, ±10%, Type QC, 500 VDCW	Quality Components
2440	C46	1.2 - 3.5 pF, Type 189-1-5, 850 V	Johnson
2364	*C47	.033 μF, 5%, Type 148P33392, 200 V	Sprague
2371	*C48	.047 μF, 10%, Type 148P47392, 200 V	Sprague
2770	C50A	180 μF, −10%, +100%, Type DFP, 200 VDCW	Sprague
	C50B	180 μF, −10%, +100%, Type DFP, 200 VDCW	Sprague
	C50C	20 μF, −10%, +100%, Type DFP, 200 VDCW	Sprague
	C50D	20 μF, −10%, +100%, Type DFP, 200 VDCW	Sprague
	C51	.02 μF, part of T2	Ballantine
	C52	.02 μF, part of T2	Ballantine
2096	C55	50 μF, −10%, +75%, Type 30D170A1, 15 VDCW	Sprague
2096	C56	50 μF, −10%, +75%, Type 30D170A1, 15 VDCW	Sprague
2555	C57	.02 μF, ±20%, Type Y5V, Style 841, 500 VDCW	Erie

*In some instruments the component listed may not be used at all or its value may differ from the value shown.

REPLACEMENT PARTS LIST (Continued)

REFER TO MODEL 320A CIRCUIT DIAGRAM, DWG. MD-1827L

B. L. Part No.	*Circuit Symbol*	*Capacitors*	*Manufacturer*
2772	C60A	30 μF, −10%, +50%, Type DFP, 450 VDCW	Sprague
	C60B	30 μF, −10%, +50%, Type DFP, 450 VDCW	Sprague
	C60C	20 μF, −10%, +50%, Type DFP, 450 VDCW	Sprague
	C60D	20 μF, −10%, +50%, Type DFP, 450 VDCW	Sprague
2792	C61A	1500 μF, −10%, +150%, Type DFP, 15 VDCW	Sprague
	C61B	1500 μF, −10%, +150%, Type DFP, 15 VDCW	Sprague
2020	C62	3000 μF, −10%, +150%, Type DFP, 10 VDCW	Sprague
2555	C70	.02 μF, ±20%, Type Y5V, Style 841, 500 VDCW	Erie
2555	C71	.02 μF, ±20%, Type Y5V, Style 841, 500 VDCW	Erie
		Resistors	
1276	R1	5.6 ohms, 5%, Type GB	Allen-Bradley
1108	R2	180 ohms, 5%, Type EB	Allen-Bradley
1072	R3	39 ohms, 5%, Type EB	Allen-Bradley
2728	R4(1)	10,000,000 ohms, 1%, Type PT-70	Pyrofilm
1015	R5	470,000 ohms, 5%, Type EB	Allen-Bradley
2729	R6(1)	337,600 ohms, 1%, Type PT-500	Pyrofilm
2730	R7(1)	9,703 ohms, 1%, Type PT-500	Pyrofilm
2731	R8(1)	316.5 ohms, 1%, Type PT-500	Pyrofilm
1463	R15	10,000,000 ohms, 1%, Type CPX-1	Aerovox
1061	R16	510 ohms, 5%, Type EB	Allen-Bradley
1061	R17	510 ohms, 5%, Type EB	Allen-Bradley
1687	R18	50,000 ohms, 1%, Type N-20	Corning
1054	R19	2,000 ohms, 5%, Type EB	Allen-Bradley
2734	R20	15 ohms, 1%, Type CPX-½	Aerovox
1040	R21	20,000 ohms, 5%, Type EB	Allen-Bradley
1010	R22	1,000,000 ohms, 5%, Type EB	Allen-Bradley
1246	R23	220 ohms, 10%, Type CB	Allen-Bradley
1034	R24	39,000 ohms, 5%, Type EB	Allen-Bradley
2620	R25	21 ohms, 1%, Type CPX-½	Aerovox
1690	R26	20,000 ohms, 5%, Type C-42S	Corning
1686	R27	3,900 ohms, 5%, Type C-20	Corning
1064	R28	220 ohms, 5%, Type EB	Allen-Bradley
1935	R29	250 ohms, 30%, Type UPM-45	Chicago Telephone
1036	R30	33,000 ohms, 5%, Type EB	Allen-Bradley
1012	R31	750,000 ohms, 5%, Type EB	Allen-Bradley
1246	R32	220 ohms, 10%, Type CB	Allen-Bradley
1034	R33	39,000 ohms, 5%, Type EB	Allen-Bradley
1690	R34	20,000 ohms, 5%, Type C-42S	Corning
1330	R35	3,000 ohms, 1%, Type CPX-½	Aerovox
1066	R36	160 ohms, 5%, Type EB	Allen-Bradley
1041	R37	18,000 ohms, 5%, Type EB	Allen-Bradley
1040	R38	20,000 ohms, 5%, Type EB	Allen-Bradley
1327	R40(2)	5,000 ohms, ±1%, Type CPX-½	Aerovox
2732	R41(2)	1,580 ohms, ±1%, Type CPX-½	Aerovox
2733	R42(2)	731 ohms, ±1%, Type CPX-½	Aerovox
1061	R45	510 ohms, 5%, Type EB	Allen-Bradley

(1) Resistors R4, R6, R7, R8 are part of a matched set and must be purchased as Set No. 1571.
(2) Resistors R40, R41, R42 are part of a matched set and must be ordered as Part No. 1568.

REPLACEMENT PARTS LIST (Continued)

REFER TO MODEL 320A CIRCUIT DIAGRAM, DWG. MD-1827L

B. L. Part No.	*Circuit Symbol*	*Resistors*	*Manufacturer*
1007	R46	2,000,000 ohms, 5%, Type EB	Allen-Bradley
1358	R47	60 ohms, 1%, Type CPX-½	Aerovox
1003	*R48	5,100,000 ohms, 5%, Type EB	Allen-Bradley
1688	R49	25,000 ohms, 2%, Type C-32	Corning
1054	R50	2,000 ohms, 5%, Type EB	Allen-Bradley
1749	R51	100 ohms, 1%, Type RLS-1B	Dahlohm
1704	R52	156 ohms, 1%, Type RLS-1B	Dahlohm
1015	R53	470,000 ohms, 5%, Type EB	Allen-Bradley
1015	R54	470,000 ohms, 5%, Type EB	Allen-Bradley
1085	R55	120,000 ohms, 5%, Type EB	Allen-Bradley
1246	R56	220 ohms, 10%, Type CB	Allen-Bradley
1246	R57	220 ohms, 10%, Type CB	Allen-Bradley
1684	R58	6,000 ohms, 1%, Type N-20	Corning
1684	R59	6,000 ohms, 1%, Type N-20	Corning
1060	R60	620 ohms, 5%, Type EB	Allen-Bradley
1060	R61	620 ohms, 5%, Type EB	Allen-Bradley
1692	R62	3,600 ohms, 5%, Type C-42S	Corning
1685	R63	27,000 ohms, 5%, Type C-20	Corning
1685	R64	27,000 ohms, 5%, Type C-20	Corning
1680	R65	20,000 ohms, 1%, Type N-20	Corning
1680	R66	20,000 ohms, 1%, Type N-20	Corning
1683	R67	600 ohms, 1%, Type N-20	Corning
1711	R68	10,000 ohms, 1%, Type RS-5	Dahlohm
1691	R69	10,000 ohms, 5%, Type C-42S	Corning
1100	R70	1,200 ohms, 5%, Type EB	Allen-Bradley
1831	R71	15,000 ohms, 10%, Type 43-15 WW Pot	Clarostat
1100	R72	1,200 ohms, 5%, Type EB	Allen-Bradley
1121	*R73	5,600 ohms, 5%, Type EB	Allen-Bradley
1003	*R74	5,100,000 ohms, 5%, Type EB	Allen-Bradley
1036	R75	33,000 ohms, 5%, Type EB	Allen-Bradley
1682	R76	1,000 ohms, 1%, Type N-20	Corning
1934	R77	400 ohms, 20%, Type UPM-45	Chicago Telephone
1681	R78	4,700 ohms, 1%, Type N-20	Corning
1680	R79	20,000 ohms, 1%, Type N-20	Corning
1057	R80	1,000 ohms, 5%, Type EB	Allen-Bradley
7257	R81	100,000 ohms, 1%, Type MF7C	Electra
7250	R82	10,400,000 ohms, 1%, Type MF72C	Electra
7251	R83	7,230,000 ohms, 1%, Type MF72C	Electra
7252	R84	3,480,000 ohms, 1%, Type MF7C	Electra
7253	R85	1,550,000 ohms, 1%, Type MF7C	Electra
7254	R86	762,000 ohms, 1%, Type MF7C	Electra
7255	R87	341,000 ohms, 1%, Type MF7C	Electra
7256	R88	138,000 ohms, 1%, Type MF7C	Electra
7262	R89	40,500 ohms, 1%, Type MF7C	Electra
1918	R90	200,000 ohms, 20%, Type UPM-45	Chicago Telephone
7259	R91	78,500 ohms, 1%, Type MF7C	Electra

*In some instruments the component listed may not be used at all or its value may differ from the value shown.

REPLACEMENT PARTS LIST (Continued)

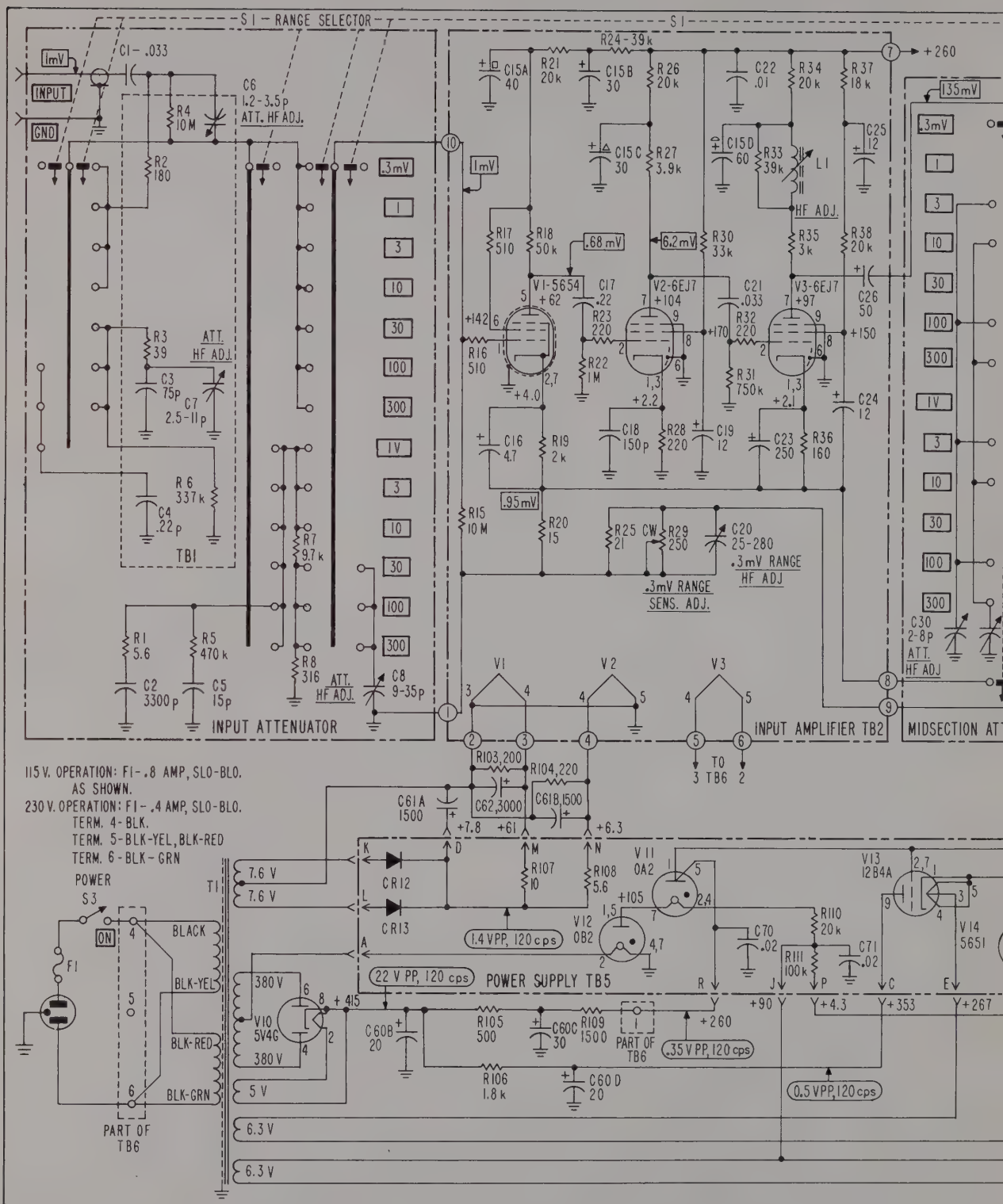
REFER TO MODEL 320A CIRCUIT DIAGRAM, DWG. MD-1827L

B. L. Part No.	*Circuit Symbol*	*Resistors*	*Manufacturer*
7258	R92	94,400 ohms, 1%, Type MF7C	Electra
7260	R93	67,500 ohms, 1%, Type MF7C	Electra
7261	R94	45,600 ohms, 1%, Type MF7C	Electra
7263	R95	34,000 ohms, 1%, Type MF7C	Electra
7264	R96	23,200 ohms, 1%, Type MF7C	Electra
7265	R97	14,600 ohms, 1%, Type MF7C	Electra
7266	R98	7,500 ohms, 1%, Type MF7C	Electra
1065	R103	200 ohms, 5%, Type EB	Allen-Bradley
1064	R104	220 ohms, 5%, Type EB	Allen-Bradley
1709	R105	500 ohms, 5%, Type 452-E-5015	Sprague
1707	R106	1,800 ohms, 5%, Type 452-E-1825	Sprague
1585	R107	10 ohms, 5%, Type BWH	IRC
6163	R108	5.6 ohms, 5%, Type BWH	IRC
1718	R109	1,500 ohms, 5%, Type 458-E-1525	Sprague
1040	R110	20,000 ohms, 5%, Type EB	Allen-Bradley
1679	R111	100,000 ohms, 1%, Type N-20	Corning
1689	R112	33,000 ohms, 5%, Type C-42S	Corning
1879	*R113	250 ohms, 5%, Type 450-E-2515	Sprague

Other Components

1579	RT1	Thermistor, 6,300 ohms, ±10%, Type 36S1	Victory Eng.
5576	CR1, 2	Diode, Germanium, Type 1N618	Amperex
5562	CR3	Diode, Silicon, Type SG22, or Type 1N816	
3144	CR4	Diode, Type S53G	Transitron
5555	CR5, 6	Diode, Type T 13G	Transitron
5555	CR7, 8	Diode, Type T 13G, or Diode, Type T 14G	Transitron
5563	CR9, 10	Diode, Type T 14G, or Diode, Type T 13G	Transitron
5563	CR11	Diode, Type T 14G, or Diode, Type T 13G	Transitron
7921	CR12, 13	Rectifier, 500 mA, Type 36591	RCA
2804	F1	Fuse, 1.2 amp, Type MDL Slo-Blo	Bussmann
3007	L1	Coil, Special	Ballantine
3068	L5	Inductor, Special	Ballantine
3190	M1	Meter,Taut Band, Special	Ballantine
3061	T1	Transformer, Power	Ballantine
3060	T2	Transformer, Output	Ballantine
3126	V1, 4	Tube, Type 5654	General Electric
5593	V2, 3, 5, 6	Tube, Type 6EJ7	Amperex
5592	V7	Tube, Type 8223, or 6DJ8	Siemens or Amperex
3138	V8, 9, 15	Lamp, Type NE-2	General Electric
3148	V10	Tube, Type 5V4G	RCA or General Electric
3106	V11	Tube, Type OA2	RCA
3128	V12	Tube, Type OB2	RCA
5569	V13	Tube, Type 12B4A	RCA
3117	V14	Tube, Type 5651	Raytheon

*In some instruments the component listed may not be used at all or its value may differ from the value shown.


S1 - RANGE SELECTOR
S1
INPUT ATTENUATOR
INPUT AMPLIFIER TB2
MIDSECTION AT
TB1
1mV
INPUT
GND
C1 - .033
C6 1.2-3.5p ATT. HF ADJ.
R4 10M
R2 180
R3 39
ATT. HF ADJ.
C3 75p
C7 2.5-11p
R6 337k
C4 .22p
R7 9.7k
R1 5.6
R5 470k
C2 3300p
C5 15p
R8 316
ATT. HF ADJ.
C8 9-35p
.3mV
1
3
10
30
100
300
1V
R24-39k
C15A 40
R21 20k
C15B 30
R26 20k
C22 .01
R34 20k
R37 18k
+260
C15C 30
R27 3.9k
C15D 60
R33 39k
L1
HF ADJ.
C25 12
R35 3k
R38 20k
.68mV
6.2mV
R30 33k
R17 510
R18 50k
V1-5654
V2-6EJ7
V3-6EJ7
C17 .22
C21 .033
C26 50
+142
+62
+104
+97
+170
+150
R23 220
R32 220
R16 510
R22 1M
R31 750k
+4.0
+2.2
+2.1
C24 12
C16 4.7
R19 2k
C18 150p
R28 220
C19 12
C23 250
R36 160
.95mV
R15 10M
R20 15
R25 CW 21
R29 250
C20 25-280
.3mV RANGE HF ADJ
.3mV RANGE SENS. ADJ.
135mV
C30 2-8p ATT. HF ADJ
V1
V2
V3
TO 3 TB6 2
R103,200
R104,220
C61A 1500
C62,3000
C61B,1500
+7.8
+61
+6.3
115 V. OPERATION: F1 - .8 AMP, SLO-BLO. AS SHOWN.
230 V. OPERATION: F1 - .4 AMP, SLO-BLO.
TERM. 4 - BLK.
TERM. 5 - BLK-YEL, BLK-RED
TERM. 6 - BLK - GRN
POWER
S3
ON
F1
T1
BLACK
BLK-YEL
BLK-RED
BLK-GRN
PART OF TB6
7.6 V
7.6 V
CR12
CR13
R107 10
R108 5.6
V11 0A2
V12 0B2
+105
1.4 VPP, 120 cps
V13 12B4A
V14 5651
C70 .02
R110 20k
R111 100k
C71 .02
380 V
380 V
V10 5V4G
+415
22 V PP, 120 cps
POWER SUPPLY TB5
C60B 20
R105 500
C60C 30
R109 1500
PART OF TB6
.35 V PP, 120 cps
+260
R106 1.8k
C60D 20
0.5 VPP, 120 cps
+90
+4.3
+353
+267
5 V
6.3 V
6.3 V

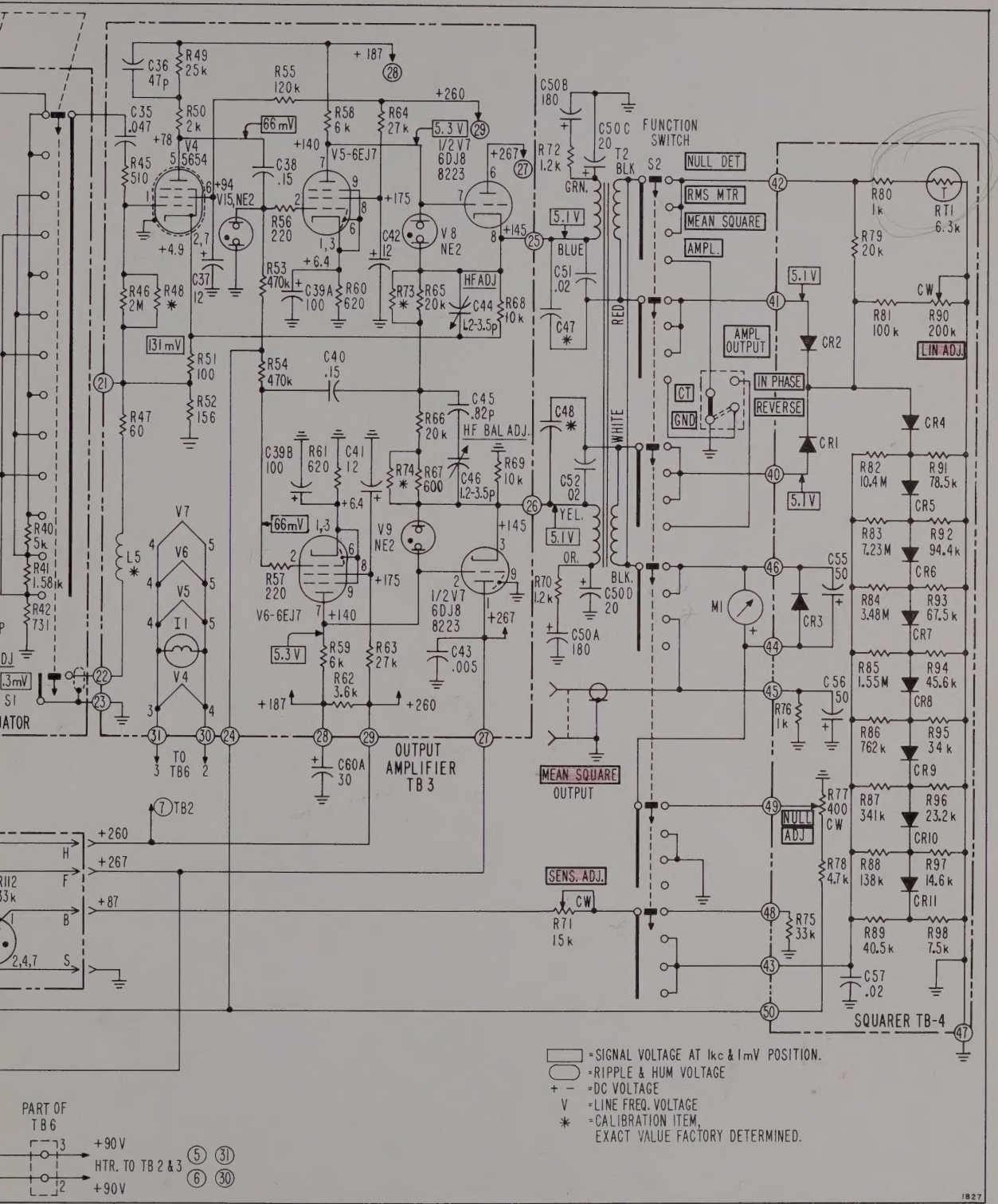


Fig. 16. Schematic of Model 320A True RMS Voltmeter
Drawing No. MD-1827L

BALLANTINE LABORATORIES, INC.
Boonton, N. J.